

**Monolithic Linear IC** 

# PWM Constant-Current Control Stepper Motor Driver

http://onsemi.com

#### Overview

The LV8732V is a 2-channel H-bridge driver IC that can switch a stepper motor driver, which supports micro-step drive with 1/8-step resolution, and two channels of a brushed motor driver, which supports forward, reverse, brake, and standby of a motor. It is ideally suited for driving brushed DC motors and stepper motors used in office equipment and amusement applications.

#### **Function**

- Single-channel PWM current control stepper motor driver (selectable with DC motor driver channel 2) incorporated.
- BiCDMOS process IC
- Low on resistance (upper side :  $0.3\Omega$ ; lower side :  $0.25\Omega$ ; total of upper and lower :  $0.55\Omega$ ; Ta = 25°C, IO = 2A)
- Micro-step mode can be set to Full-step, Half-step, Quarter-step, or 1/8-step
- Excitation step proceeds only by step signal input
- Motor current selectable in four steps
- Output short-circuit protection circuit (selectable from latch-type or auto-reset-type) incorporated
- Unusual condition warning output pins
- No control power supply required

#### **Specifications**

#### **Absolute Maximum Ratings** at Ta = 25°C

Parameter	Symbol	Conditions	Ratings	Unit
Supply voltage	VM max	VM , VM1 , VM2	36	V
Output peak current	I <sub>O</sub> peak	tw ≤ 10ms , duty 20% , Per 1ch	2.5	Α
Output current	I <sub>O</sub> max	Per 1ch	2	Α
Logic input voltage	VIN	ATT1, ATT2, EMM, RST/BLK, STEP/DC22 FR/DC21, MD2/DC12, MD1/DC11, DM, OE, ST	-0.3 to +6	V
MONI/EMO input voltage	Vmoni/Vemo		-0.3 to +6	V
Allowable power dissipation	Pd max	*	3.25	W
Operating temperature	Topr		-20 to +85	°C
Storage temperature	Tstg		-55 to +150	°C

<sup>\*</sup> Specified circuit board : 90.0mm×90.0mm×1.6mm, glass epoxy 2-layer board, with backside mounting.

Caution 1) Absolute maximum ratings represent the value which cannot be exceeded for any length of time.

Caution 2) Even when the device is used within the range of absolute maximum ratings, as a result of continuous usage under high temperature, high current, high voltage, or drastic temperature change, the reliability of the IC may be degraded. Please contact us for the further details.

Stresses exceeding Maximum Ratings may damage the device. Maximum Ratings are stress ratings only. Functional operation above the Recommended Operating Conditions is not implied. Extended exposure to stresses above the Recommended Operating Conditions may affect device reliability.

#### Allowable Operating Ratings at Ta = 25°C

Parameter	Symbol	Conditions	Ratings	Unit
Supply voltage range	VM	VM , VM1 , VM2	9 to 32	V
Logic input voltage	VIN	ATT1 , ATT2 , EMM , RST/BLK , STEP/DC22 , FR/DC21 , MD2/DC12 , MD1/DC11 , DM , OE , ST	0 to 5.5	V
VREF input voltage range	VREF		0 to 3	V

#### ORDERING INFORMATION

See detailed ordering and shipping information on page 26 of this data sheet.

# **Electrical Characteristics** at Ta = 25°C, VM = 24V, VREF = 1.5V

Paramet	ter	Symbol	Conditions		Ratings		Unit
				min	typ	max	
Standby mode current drain		IMst	ST = "L" , I(VM)+I(VM1)+I(VM2)		100	400	μА
Current drain		IM	ST = "H", OE = "L", with no load ,		3.2	5	mA
VREG5 output voltage	10	Vreg5	$I(VM)+I(VM1)+I(VM2)$ $I_{O} = -1m\Delta$	4.5	5	5.5	V
			I <sub>O</sub> = -1mA	+			°C
Thermal shutdown to	•	TSD	Design guarantee	150	180	200	
Thermal hysteresis w	viatn	ΔTSD	Design guarantee		40		°C
Motor driver		I p	L. OA Harrista and statement	1	0.0	0.4	
Output on resistance		Ronu	I <sub>O</sub> = 2A, Upper-side on resistance		0.3	0.4	Ω
0.1: 11::1::::::::::::::::::::::::::::::	-1	Rond	I <sub>O</sub> = 2A, Lower-side on resistance		0.25	0.33	Ω
Output leakage curre		I <sub>O</sub> leak				50	μA
Diode forward voltag		VD	ID = -2A		1.2	1.4	V
Logic pin input currer	nt	I <sub>IN</sub> L	ATT1 , ATT2 , EMM , RST/BLK , STEP/DC22 , FR/DC21 , MD2/DC12 , MD1/DC11 , DM , OE , ST , V <sub>IN</sub> = 0.8V	4	8	12	μА
		I <sub>IN</sub> H	V <sub>IN</sub> = 5V	30	50	70	μА
Logic input voltage	High	V <sub>IN</sub> h	ATT1, ATT2, EMM, RST/BLK,	2.0		5.5	V
	Low	V <sub>IN</sub> I	STEP/DC22 , FR/DC21 , MD2/DC12 , MD1/DC11 , DM , OE , ST	0		0.8	V
Current setting comparator	1/8 step resolution	Vtdac0_2W	Step 0 (When initialized : channel 1 comparator level)	0.291	0.3	0.309	V
threshold voltage		Vtdac1_2W	Step 1 (Initial state+1)	0.285	0.294	0.303	V
(current step		Vtdac2_2W	Step 2 (Initial state+2)	0.267	0.276	0.285	V
switching)		Vtdac3_2W	Step 3 (Initial state+3)	0.240	0.249	0.258	V
		Vtdac4 2W	Step 4 (Initial state+4)	0.201	0.21	0.219	V
		Vtdac5 2W	Step 5 (Initial state+5)	0.157	0.165	0.173	V
		Vtdac6 2W	Step 6 (Initial state+6)	0.107	0.114	0.121	V
		Vtdac7_2W	Step 7 (Initial state+7)	0.053	0.06	0.067	V
	Quarter step	Vtdac0_W	Step 0 (When initialized : channel 1	0.291	0.3	0.309	V
	resolution	Vtdac2_W	comparator level) Step 2 (Initial state+1)	0.267	0.276	0.285	V
		Vtdac2_vv Vtdac4_W	Step 4 (Initial state+2)	0.201	0.21	0.219	
		Vtdac4_VV Vtdac6 W	Step 6 (Initial state+3)	0.201	0.21	0.121	
	Half step	Vtdac0_vv	Step 0 (When initialized : channel 1	0.107	0.114	0.309	
	resolution	_	comparator level)				
		Vtdac4_H	Step 4 (Initial state+1)	0.201	0.21	0.219	V
	Full step resolution	Vtdac4_F	Step 4' (When initialized : channel 1 comparator level)	0.291	0.3	0.309	V
Current setting comp	parator	Vtatt00	ATT1 = L, ATT2 = L	0.291	0.3	0.309	V
threshold voltage		Vtatt01	ATT1 = H, ATT2 = L	0.232	0.24	0.248	V
(current attenuation r	rate switching)	Vtatt10	ATT1 = L, ATT2 = H	0.143	0.15	0.157	V
		Vtatt11	ATT1 = H, ATT2 = H	0.053	0.06	0.067	V
Chopping frequency		Fchop	Cchop = 200pF	40	50	60	kHz
CHOP pin charge/dis	scharge current	Ichop		7	10	13	μΑ
Chopping oscillation circuit		Vtup		0.8	1	1.2	V
threshold voltage		Vtdown		0.4	0.5	0.6	
· · ·		Iref	VREF = 1.5V	-0.5			μА
MONI pin saturation	voltage	Vsatmon	Imoni = 1mA			400	mV
Charge pump		1	T	<del>, , , , , , , , , , , , , , , , , , , </del>	П	1	
VG output voltage		VG		28	28.7	29.8	V
Rise time		tONG	$VG = 0.1 \mu F$ , Between CP1-CP2 0.1u F , ST="H" $\rightarrow$ VG=VM+4V		200	500	μS
Oscillator frequency		Fosc		90	125	150	kHz

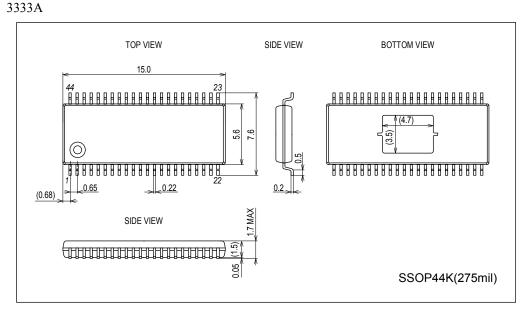
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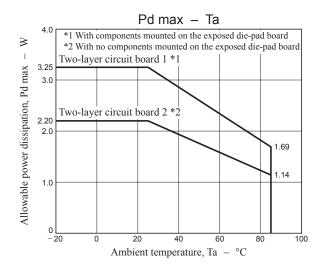
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Description	O. made al	Conditions	Ratings			11.2
Parameter	Symbol	Conditions	min	typ	max	Unit
Output short-circuit protection						
EMO pin saturation voltage	Vsatemo	Iemo = 1mA			400	mV
CEM pin charge current	Icem	Vcem = 0V	7	10	13	μА
CEM pin threshold voltage	Vtcem		0.8	1	1.2	V

# **Package Dimensions**

unit: mm (typ)



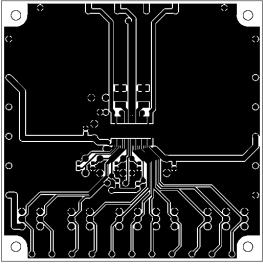


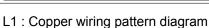
#### Substrate Specifications (Substrate recommended for operation of LV8732V)

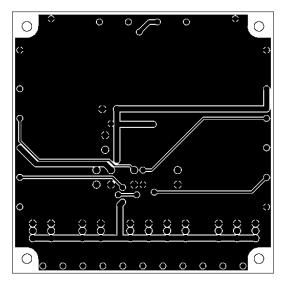
Size :  $90\text{mm} \times 90\text{mm} \times 1.6\text{mm}$  (two-layer substrate [2S0P])

Material : Glass epoxy

Copper wiring density : L1 = 85% / L2 = 90%







L2: Copper wiring pattern diagram

#### **Cautions**

- 1) The data for the case with the Exposed Die-Pad substrate mounted shows the values when 90% or more of the Exposed Die-Pad is wet.
- 2) For the set design, employ the derating design with sufficient margin.

Stresses to be derated include the voltage, current, junction temperature, power loss, and mechanical stresses such as vibration, impact, and tension.

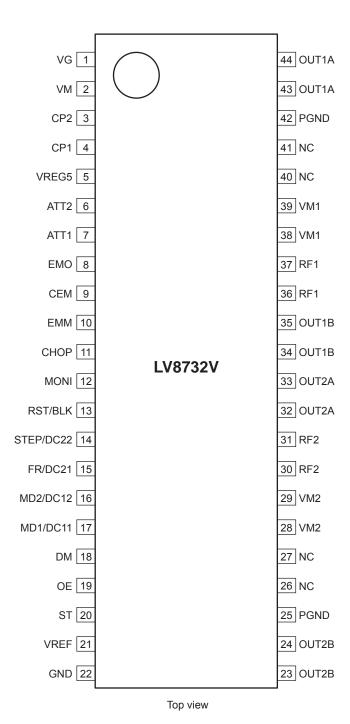
Accordingly, the design must ensure these stresses to be as low or small as possible.

The guideline for ordinary derating is shown below:

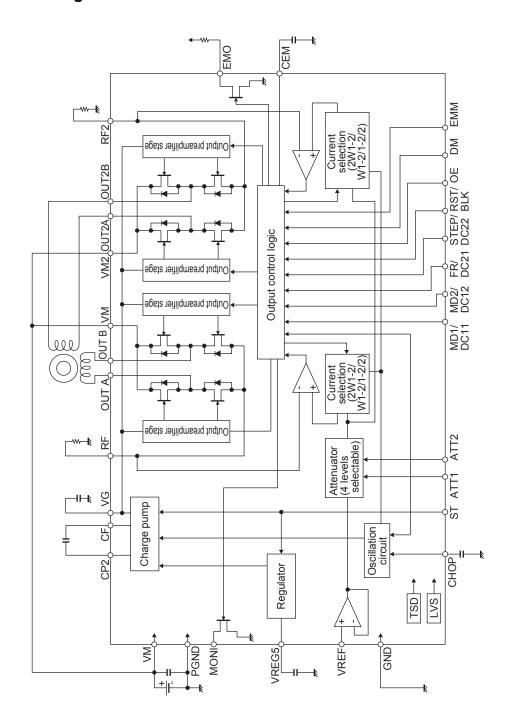
- (1)Maximum value 80% or less for the voltage rating
- (2)Maximum value 80% or less for the current rating
- (3)Maximum value 80% or less for the temperature rating
- 3) After the set design, be sure to verify the design with the actual product.

Confirm the solder joint state and verify also the reliability of solder joint for the Exposed Die-Pad, etc. Any void or deterioration, if observed in the solder joint of these parts, causes deteriorated thermal conduction, possibly resulting in thermal destruction of IC.

# **Pin Assignment**



# **Block Diagram**



#### **Pin Functions**

PINFU	inctions		
Pin No.	Pin Name	Pin Function	Equivalent Circuit
6	ATT2	Motor holding current switching pin.	
7	ATT1	Motor holding current switching pin.	
10	EMM	Output short-circuit protection mode	
		switching pin.	VREG5 ○ •
13	RST/BLK	RESET input pin (STM) / Blanking time	
		switching pin (DCM).	<b>★</b>
14	STEP/DC22	STEP signal input pin (STM) / Channel 2	
	0.2.75022	output control input pin 2 (DCM).	<b>★</b> • • • • • • • • • • • • • • • • • • •
15	FR/DC21	CW / CCW signal input pin (STM) /	1,0kΩ
10	1100021	Channel 2 output control input pin 1	10K22 W
		(DCM).	
16	MD2/DC12	Excitation mode switching pin 2 (STM) /	
10	NID2/DC12	Channel 1 output control input pin 2	★ \$100kΩ ↑
47	MD4/DO44	(DCM).	
17	MD1/DC11	Excitation mode switching pin 1 (STM) /	
		Channel 1 output control input pin 1	GND O + +
40	DM	(DCM).	
18	DM	Drive mode (STM/DCM) switching pin.	
19	OE	Output enable signal input pin.	
20	ST	Chip enable pin.	VREG5 ○
			<b>*</b>
			₩
			ξ20kΩ
			10kΩ 
			<b>↓</b>   · · · · · · · · · · · · · · · · · ·
			<b>↑</b> [
			\$80kΩ
			GND O
23, 24	OUT2B	Channel 2 OUTB output pin.	38(39)
25, 42	PGND	Power system ground.	28/29
28, 29	VM2	Channel 2 motor power supply	
		connection pin.	
30, 31	RF2	Channel 2 current-sense resistor	
		connection pin.	
32, 33	OUT2A	Channel 2 OUTA output pin.	
34, 35	OUT1B	Channel 1 OUTB output pin.	
36, 37	RF1	Channel 1 current-sense resistor	(43)44) (34)35)
		connection pin.	(23)24)
38, 39	VM1	Channel 1 motor power supply pin.	│
43, 44	OUT1A	Channel 1 OUTA output pin.	
			1010
			10κΩ 500Ω
			25/42 N 500Ω μ
			(25)(42)
			36(37)
			30(31)
			GND O

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Continued from preceding page. Pin No. Pin Name Pin Function Equivalent Circuit 1 ۷G Charge pump capacitor connection pin. (3) (4) (1) 2 VM Motor power supply connection pin. VREG5 O-3 CP2 Charge pump capacitor connection pin. Charge pump capacitor connection pin. 4 CP1 GND C 21 VREF Constant current control reference VREG5 Ovoltage input pin.  $500\Omega$ GND ○ 5 VREG5 Internal power supply capacitor VM Oconnection pin. **≷**78kΩ ≶26kΩ GND O-8 EMO Output short-circuit state warning output VREG5 ○ 12 MONI Position detection monitor pin. GND O-

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	from preceding	page.	
Pin No.	Pin Name	Pin Function	Equivalent Circuit
9	CEM	Pin to connect the output short-circuit state detection time setting capacitor.	VREG5 O SOUD SOUD SOUD SOUD SOUD SOUD SOUD SO
11	СНОР	Chopping frequency setting capacitor connection pin.	VREG5 Ο 500Ω 500Ω
22	GND	Ground.	
26, 27	NC	No Connection	
40, 41		(No internal connection to the IC)	

#### **Description of operation**

#### 1.Input Pin Function

#### 1-1) Chip enable function

This IC is switched between standby and operating mode by setting the ST pin. In standby mode, the IC is set to power-save mode and all logic is reset. In addition, the internal regulator circuit and charge pump circuit do not operate in standby mode.

l	ST	Mode	Internal regulator	Charge pump
	Low or Open	Standby mode	Standby	Standby
	High	Operating mode	Operating	Operating

#### 1-2) Drive mode switching pin function

The IC drive mode is switched by setting the DM pin. In STM mode, stepper motor channel 1 can be controlled by the CLK-IN input. In DCM mode, DC motor channel 2 or stepper motor channel 1 can be controlled by parallel input. Stepper motor control using parallel input is Full-step or Half-step full torque.

DM	Drive mode	Application
Low or Open	STM mode	Stepper motor channel 1 (CLK-IN)
High	DCM mode	DC motor channel 2 or stepper motor channel 1 (parallel)

#### 2.STM mode (DM = Low or Open)

#### 2-1) STEP pin function

In	put	Operating mode
	pul	Operating mode
ST	STP	
Low	*	Standby mode
High		Excitation step proceeds
High		Excitation step is kept

#### 2-2) Excitation mode setting function

MD1	MD2 Micro-step resolution Initial position			position
		(Excitation mode)	Channel 1	Channel 2
Low	Low	Full step(2 phase excitation)	100%	-100%
High	Low	Half step(1-2 phase excitation)	100%	0%
Low	High	Quarter step (W1-2 phase excitation)	100%	0%
High	High	1/8 step(2W1-2 phase excitation)	100%	0%

This is the initial position of each excitation mode in the initial state after power-on and when the counter is reset.

#### 2-3) Position detection monitoring function

The MONI position detection monitoring pin is of an open drain type.

When the excitation position is in the initial position, the MONI output is placed in the ON state.

(Refer to "2-12.Examples of current waveforms in each micro-step mode .")

#### 2-4) Setting constant-current control reference current

This IC is designed to automatically exercise PWM constant-current chopping control for the motor current by setting the output current. Based on the voltage input to the VREF pin and the resistance connected between RF and GND, the output current that is subject to the constant-current control is set using the calculation formula below:

$$I_{OUT} = (VREF/5)/RF$$
 resistance

The voltage input to the VREF pin can be switched to four-step settings depending on the statuses of the two inputs, ATT1 and ATT2. This is effective for reducing power consumption when motor holding current is supplied.

Attenuation function for VREF input voltage

ATT1	ATT2	Current setting reference voltage attenuation ratio
Low	Low	100%
High	Low	80%
Low	High	50%
High	High	20%

The formula used to calculate the output current when using the function for attenuating the VREF input voltage is given below.

$$I_{OUT} = (VREF/5) \times (attenuation ratio)/RF resistance$$

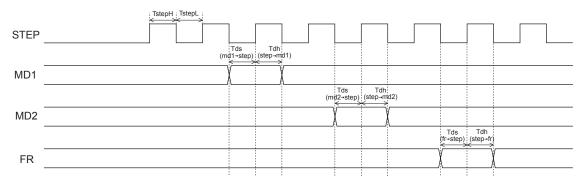
Example : At VREF of 1.5V, a reference voltage setting of 100% [(ATT1, ATT2) = (L, L)] and an RF resistance of  $0.3\Omega$ , the output current is set as shown below.

$$I_{OUT} = 1.5V/5 \times 100\%/0.3\Omega = 1.0A$$

If, in this state, (ATT1, ATT2) is set to (H, H), IOUT will be as follows : 
$$I_{OUT} = 1.0A \times 20\% = 200mA$$

In this way, the output current is attenuated when the motor holding current is supplied so that power can be conserved.

#### 2-5) Input Timing



TstepH/TstepL : Clock H/L pulse width (min 500ns)

Tds: Data set-up time (min 500ns)
Tdh: Data hold time (min 500ns)

#### 2-6) Blanking period

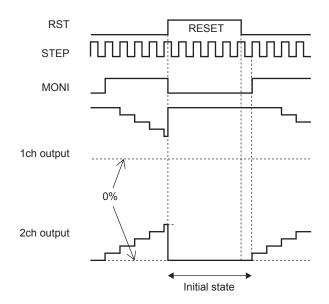
If, when exercising PWM constant-current chopping control over the motor current, the mode is switched from decay to charge, the recovery current of the parasitic diode may flow to the current sensing resistance, causing noise to be carried on the current sensing resistance pin, and this may result in erroneous detection. To prevent this erroneous detection, a blanking period is provided to prevent the noise occurring during mode switching from being received. During this period, the mode is not switched from charge to decay even if noise is carried on the current sensing resistance pin.

In the stepper motor driver mode (DM = Low or Open) of this IC, the blanking time is fixed at approximately  $1\mu$ s. In the DC motor driver mode (DM = High), the blanking time can be switched to one of two levels using the RST/BLK pin. (Refer to "Blanking time switching function.")

<sup>\*</sup> The above setting is the output current at 100% of each excitation mode.

2-7) Reset function

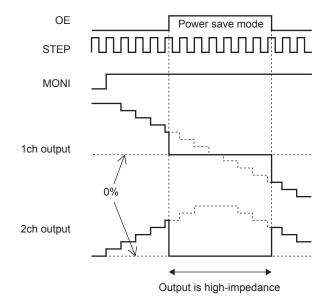
RST	Operating mode
Low	Normal operation
High	Reset state



When the RST pin is set to High, the excitation position of the output is forcibly set to the initial state, and the MONI output is placed in the ON state. When RST is then set to Low, the excitation position is advanced by the next STEP input.

2-8) Output enable function

OE	Operating mode	
Low	Output ON	
High	Output OFF	

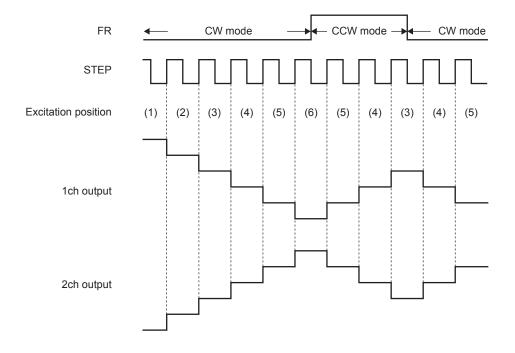


When the OE pin is set High, the output is forced OFF and goes to high impedance.

However, the internal logic circuits are operating, so the excitation position proceeds when the STEP signal is input. Therefore, when OE is returned to Low, the output level conforms to the excitation position proceeded by the STEP input.

2-9) Forward/reverse switching function

FR	Operating mode		
Low	Clockwise (CW)		
High	Counter-clockwise (CCW)		



The internal D/A converter proceeds by one bit at the rising edge of the input STEP pulse.

In addition, CW and CCW mode are switched by setting the FR pin.

In CW mode, the channel 2 current phase is delayed by 90° relative to the channel 1 current.

In CCW mode, the channel 2 current phase is advanced by 90° relative to the channel 1 current.

#### 2-10) Chopping frequency setting

For constant-current control, this IC performs chopping operations at the frequency determined by the capacitor (Cchop) connected between the CHOP pin and GND.

The chopping frequency is set as shown below by the capacitor (Cchop) connected between the CHOP pin and GND.

Fchop = Ichop/ (Cchop 
$$\times$$
 Vtchop  $\times$  2) (Hz)

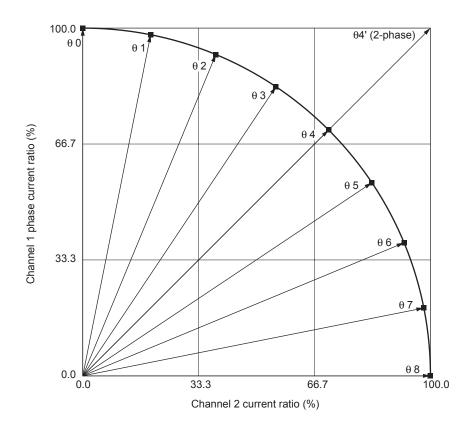
Ichop : Capacitor charge/discharge current, typ  $10\mu A$ 

Vtchop: Charge/discharge hysteresis voltage (Vtup-Vtdown), typ 0.5V

For instance, when Cchop is 200pF, the chopping frequency will be as follows:

Fchop = 
$$10\mu A/(200pF \times 0.5V \times 2) = 50kHz$$

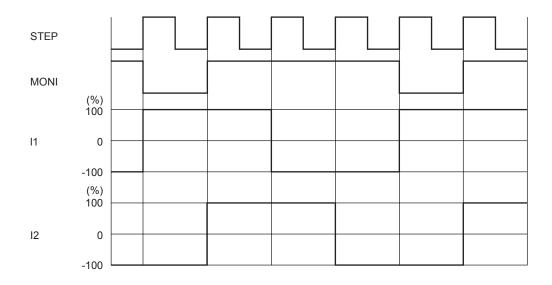
# 2-11) Output current vector locus (one step is normalized to 90 degrees)



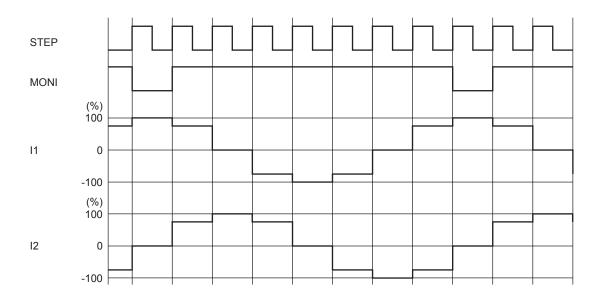
## Setting current ration in each micro-step mode

STEP	1/8 ste	ep (%)	Quarter	step (%)	Half ste	ep (%)	Full ste	p (%)
	Channel 1	Channel 2						
θ0	100	0	100	0	100	0		
θ1	98	20						
θ2	92	38	92	38				
θ3	83	55						
θ4	70	70	70	70	70	70	100	100
θ5	55	83						
θ6	38	92	38	92				
θ7	20	98						
θ8	0	100	0	100	0	100		

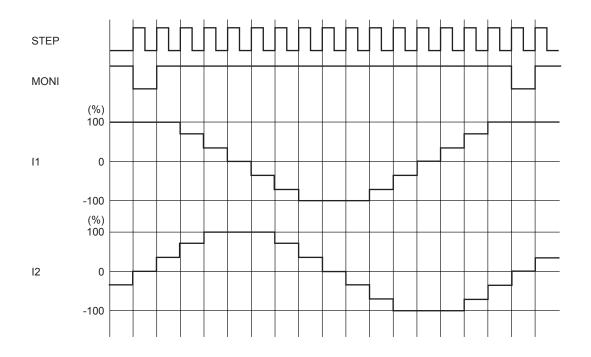
# 2-12) Examples of current waveform in each micro-step mode Full step (CW mode)



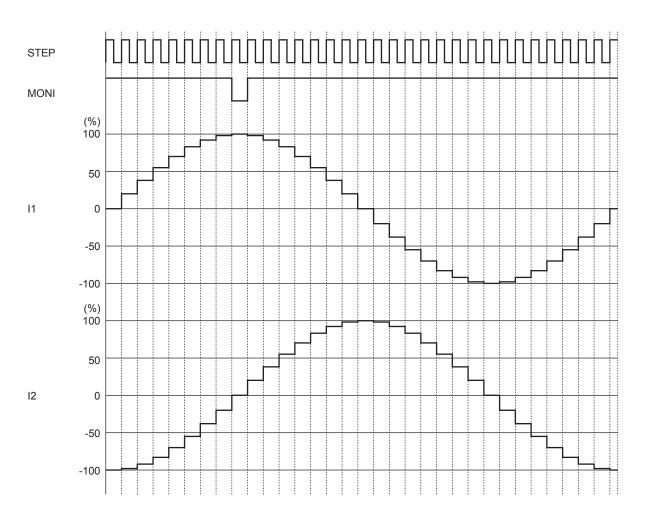
# Half step (CW mode)



Quarter step (CW mode)

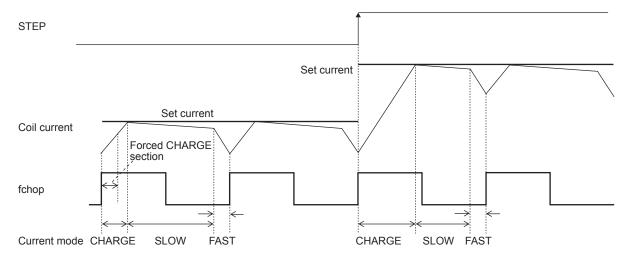


1/8 step (CW mode)

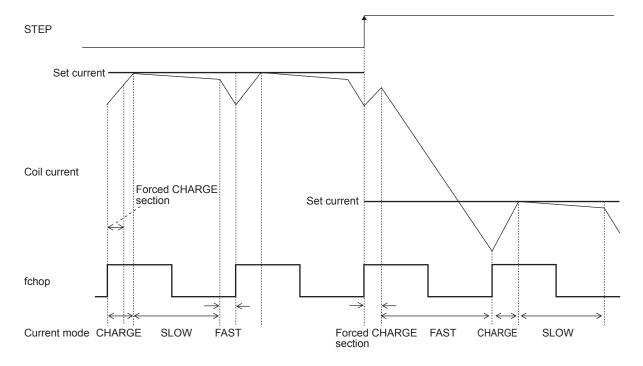


#### 2-13) Current control operation specification

(Sine wave increasing direction)



(Sine wave decreasing direction)



In each current mode, the operation sequence is as described below:

- At rise of chopping frequency, the CHARGE mode begins. (In the time defined as the "blanking time," the CHARGE mode is forced regardless of the magnitude of the coil current (ICOIL) and set current (IREF).)
- The coil current (ICOIL) and set current (IREF) are compared in this blanking time.

When (ICOIL < IREF) state exists;

The CHARGE mode up to ICOIL  $\geq$  IREF, then followed by changeover to the SLOW DECAY mode, and finally by the FAST DECAY mode for approximately 1 $\mu$ s.

When (ICOIL < IREF) state does not exist;

The FAST DECAY mode begins. The coil current is attenuated in the FAST DECAY mode till one cycle of chopping is over.

Above operations are repeated. Normally, the SLOW (+FAST) DECAY mode continues in the sine wave increasing direction, then entering the FAST DECAY mode till the current is attenuated to the set level and followed by the SLOW DECAY mode.

#### 3. DCM Mode (DM-High)

## 3-1) DCM mode output control logic

Paralle	el input	Output		Output Mode		Mode
DC11 (21)	DC12 (22)	OUT1 (2) A	OUT1 (2) B			
Low	Low	OFF	OFF	Standby		
High	Low	High	Low	CW (Forward)		
Low	High	Low	High	CCW (Reverse)		
High	High	Low	Low	Brake		

## 3-2) Blanking time switching function

BLK	Blanking time	
Low	2μs	
High	3μs	

#### 3-3) Output enable function

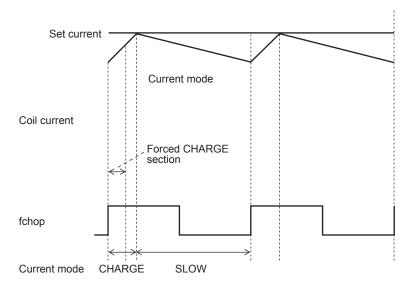
OE	Operating mode	
Low	Output ON	
High	Output OFF	

When the OE pin is set High, the output is forced OFF and goes to high impedance. When the OE pin is set Low, output conforms to the control logic.

#### 3-4) Current limit reference voltage setting function

By setting a current limit, this IC automatically exercises short braking control to ensure that when the motor current has reached this limit, the current will not exceed it.

(Current limit control time chart)



The limit current is set as calculated on the basis of the voltage input to the VREF pin and the resistance between the RF pin and GND using the formula given below.

Ilimit = 
$$(VREF/5)/RF$$
 resistance

The voltage applied to the VREF pin can be switched to any of the four setting levels depending on the statuses of the two inputs, ATT1 and ATT2.

Function for attenuating VREF input voltage

ATT1	ATT2	Current setting reference voltage attenuation ratio
Low	Low	100%
High	Low	80%
Low	High	50%
High	High	20%

The formula used to calculate the output current when using the function for attenuating the VREF input voltage is given below.

Ilimit =  $(VREF/5) \times (attenuation ratio) / RF resistance$ 

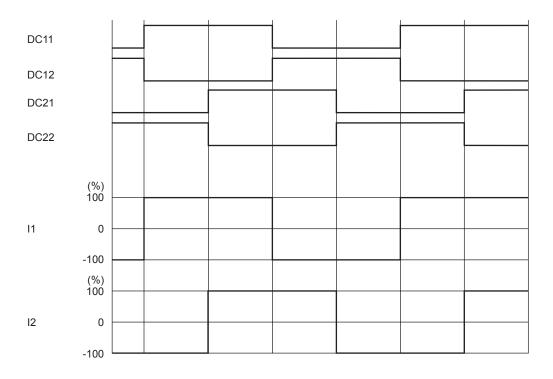
Example : At VREF of 1.5V, a reference voltage setting of 100% [(ATT1, ATT2) = (L, L)] and an RF resistance of  $0.3\Omega$ , the output current is set as shown below.

Ilimit = 
$$1.5V/5 \times 100\%/0.3\Omega = 1.0A$$

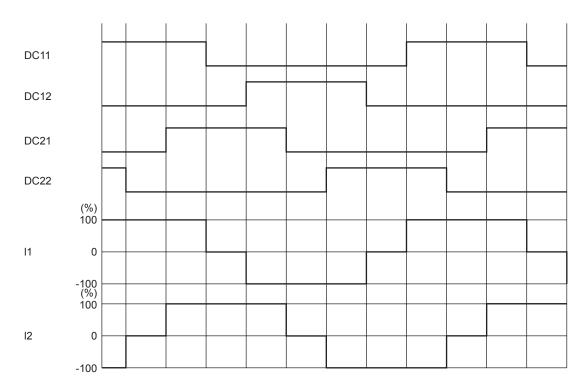
If, in this state, (ATT1, ATT2) has been set to (H, H), Ilimit will be as follows :

Ilimit = 
$$1.0A \times 20\% = 200mA$$

3-5) Examples of current waveform in each micro-step mode when stepper motor parallel input control Full step (CW mode)



Half step full torque (CW mode)



#### 4. Output short-circuit protection function

This IC incorporates an output short-circuit protection circuit that, when the output has been shorted by an event such as shorting to power or shorting to ground, sets the output to the standby mode and turns on the warning output in order to prevent the IC from being damaged. In the stepping motor driver (STM) mode (DM = Low), this function sets the output to the standby mode for both channels by detecting the short-circuiting in one of the channels. In the DC motor driver mode (DM = High), channels 1 and 2 operate independently. (Even if the output of channel 1 has been short-circuited, channel 2 will operate normally.)

#### 4-1) Output short-circuit protection mode switching function

Output short-circuit protection mode of IC can be switched by the setting of EMM pin .

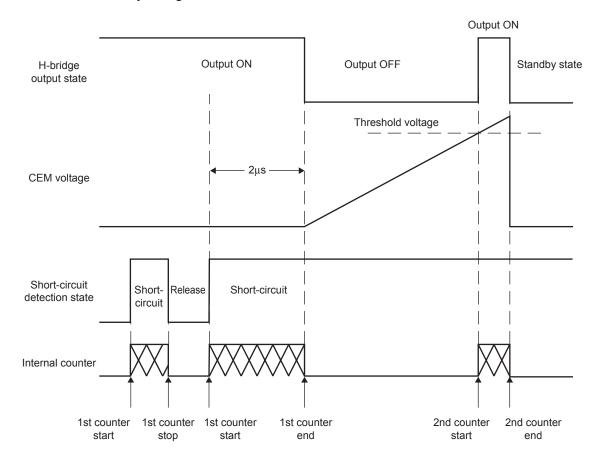
EMM	State
Low or Open	Latch method
High	Auto reset method

#### 4-2) Latch type

In the latch mode, when the output current exceeds the detection current level, the output is turned OFF, and this state is held.

The detection of the output short-circuited state by the IC causes the output short-circuit protection circuit to be activated.

When the short-circuited state continues for the period of time set using the internal timer (approximately  $2\mu s$ ), the output in which the short-circuiting has been detected is first set to OFF. After this, the output is set to ON again as soon as the timer latch time (Tcem) described later has been exceeded, and if the short-circuited state is still detected, all the outputs of the channel concerned are switched to the standby mode, and this state is held. This state is released by setting ST to low.



#### 4-3) Auto reset type

In the automatic reset mode, when the output current exceeds the detection current level, the output waveform changes to the switching waveform.

As with the latch system, when the output short-circuited state is detected, the short-circuit protection circuit is activated. When the operation of the short-circuit detection circuit exceeds the timer latch time (Tcem) described later, the output is changed over to the standby mode and is reset to the ON mode again in 2ms (typ). In this event, if the over current mode still continues, the switching mode described above is repeated until the over current mode is canceled.

#### 4-4) Unusual condition warning output pins (EMO, MONI)

The LV8731V is provided with the EMO pin which notifies the CPU of an unusual condition if the protection circuit operates by detecting an unusual condition of the IC. This pin is of the open-drain output type and when an unusual condition is detected, the EMO output is placed in the ON (EMO = Low) state.

In the DC motor driver mode (DM = High), the MONI pin also functions as a warning output pin.

The functions of the EMO pin and MONI pin change as shown below depending on the state of the DM pin.

When the DM is low (STM mode):

EMO: Unusual condition warning output pin

MONI: Excitation initial position detection monitoring

When the DM is high (DCM) mode):

EMO: Channel 1 warning output pin MONI: Channel 2 warning output pin

Furthermore, the EMO (MONI) pin is placed in the ON state when one of the following conditions occurs.

- 1. Shorting-to-power, shorting-to-ground, or shorting-to-load occurs at the output pin and the output short-circuit protection circuit is activated.
- 2. The IC junction temperature rises and the thermal protection circuit is activated.

Unusual condition	DM = L (S	TM mode)	DM = H (DCM mode)		
	EMO	MONI	EMO	MONI	
Channel 1 short-circuit detected	ON	-	ON	-	
Channel 2 short-circuit detected	ON	-	-	ON	
Overheating condition detected	ON	-	ON	ON	

#### 4-5) Timer latch time (Tcem)

The time taken for the output to be set to OFF when the output has been short-circuited can be set using capacitor Ccem, connected between the CEM pin and GND. The value of capacitor Ccem is determined by the formula given below.

Timer latch : Tcem  $\sim$  Ccem  $\times$  Vtcem/Icem [sec]

Vtcem: Comparator threshold voltage, typ 1V Icem: CEM pin charge current, typ 10μA

#### 5. Thermal shutdown function

The thermal shutdown circuit is included, and the output is turned off when junction temperature Tj exceeds 180°C and the abnormal state warning output is turned on at the same time.

When the temperature falls hysteresis level, output is driven again (automatic restoration)

The thermal shutdown circuit doesn't guarantee protection of the set and the destruction prevention of IC, because it works at the temperature that is higher than rating (Tjmax=150°C) of the junction temperature

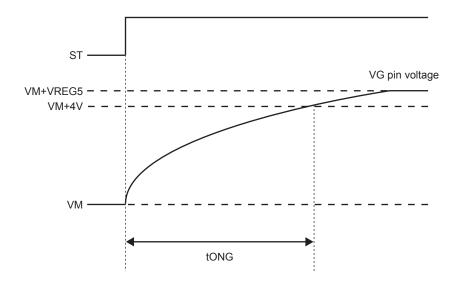
 $TSD = 180^{\circ}C \text{ (typ)}$ 

 $\Delta$ TSD = 40°C (typ)

#### 6. Charge Pump Circuit

When the ST pin is set High, the charge pump circuit operates and the VG pin voltage is boosted from the VM voltage to the VM + VREG5 voltage.

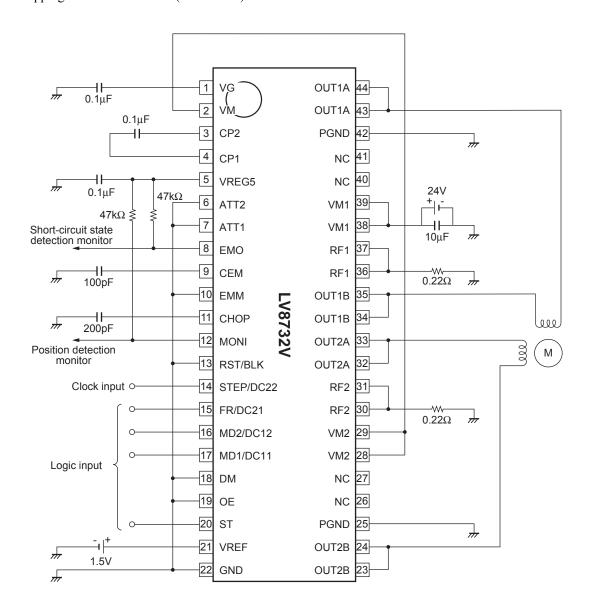
If the VG pin voltage is not boosted to VM+4V or more, the output pin cannot be turned on. Therefore it is recommended that the drive of motor is started after the time has passed tONG or more.



VG Pin Voltage Schematic View

# **Application Circuit Example**

• Stepping motor driver circuit (DM = Low)



The formulae for setting the constants in the examples of the application circuits above are as follows: Constant current (100%) setting

When 
$$VREF = 1.5V$$

$$I_{OUT} = VREF/5/RF$$
 resistance  
= 1.5V/5/0.22 $\Omega$  = 1.36A

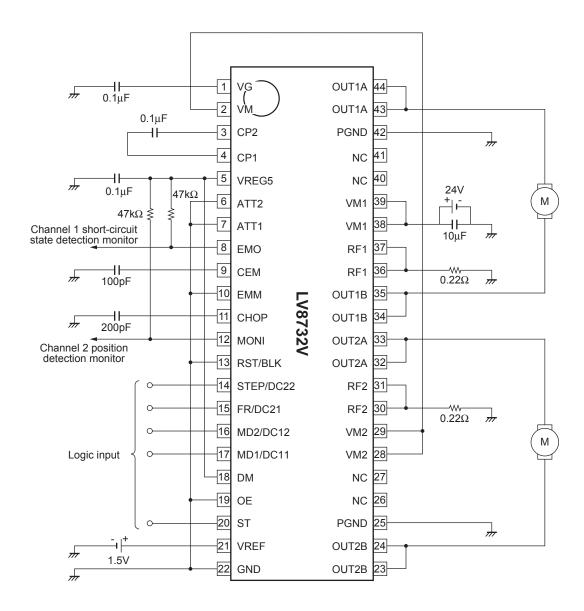
Chopping frequency setting

Fchop = Ichop/ (Cchop × Vtchop × 2)  
= 
$$10\mu$$
A/ ( $200pF \times 0.5V \times 2$ ) =  $50kHz$ 

Timer latch time when the output is short-circuited

$$Tcem = Ccem \times Vtcem/Icem$$
$$= 100pF \times 1V/10\mu A = 10\mu s$$

• DC motor driver circuit (DM = High, and the current limit function is in use.)



The formulae for setting the constants in the examples of the application circuits above are as follows: Constant current limit (100%) setting

When 
$$VREF = 1.5V$$

Ilimit = VREF/5/RF resistance  
= 
$$1.5$$
V/5/0.22 $\Omega$  = 1.36A

Chopping frequency setting

Fchop = Ichop/ (Cchop × Vtchop × 2)  
= 
$$10\mu$$
A/ (200pF × 0.5V × 2) =  $50k$ Hz

Timer latch time when the output is short-circuited

Tcem = Ccem × Vtcem/Icem  
= 
$$100 pF \times 1V/10 \mu A = 10 \mu s$$

#### ORDERING INFORMATION

Device	Package	Shipping (Qty / Packing)
LV8732V-TLM-H	SSOP44K (275mil) (Pb-Free / Halogen Free)	2000 / Tape & Reel
LV8732V-MPB-H	SSOP44K (275mil) (Pb-Free / Halogen Free)	30 / Fan-Fold

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