

AM/FM TUNER FOR CAR RADIO AND HIFI APPLICATIONS

FM-PART

- RF AGC GENERATION BY RF AND IF DETECTION FOR PIN DIODES AND MOSFET (PRESTAGE)
- 1ST MIXER FOR 1ST FM IF 10.7MHz WITH PROGRAMMABLE IF TANK ADJUST FOR FM AND AM UPCONVERSION
- 2 PROGRAMMABLE IF-GAIN STAGES
- 2ND MIXER FOR 2ND IF 450KHz
- INTERNAL IF BANDPASS FILTER WITH THREE BANDWIDTHS CONTROLLED BY ISS (INCLUDING WEATHER BAND)
- FULLY INTEGRATED FM-DEMODULATOR

AM-PART

- WIDE AND NARROW AGC GENERATION
- PREAMPLIFIER AND MIXER FOR ST 'F 10.7MHZ, AM UPCONVERSION
- 2ND MIXER FOR 2ND IF 450 (HZ
- INTEGRATED AM-DEMODULATOR
- OUTPUT FOR AM-STEREO-DECULSK

ADDITIONAL FEATURES

- EUCH PERFORMATICE FAST PLL FOR RDS-SYSTEM
- IF COUNTER FOR FM AND AM UPCON 'ERSION WITH SEARCH STOP SIGNAL
- DEVIALITY DETECTOR FOR LEVEL, DEVIATION, ADJACENT CHANNEL AND



MULTIPATH

- QUALITY DETECTION INFORMATIONS AS
- ► iSS (INTELL'GEN SELECTIVITY SYSTEM) FOR CANCELLATION OF ADJACENT CHANINEL AND NOISE INFLUENCES
- A CASENT CHANNEL MUTE
- ▲ FULLY ELECTRONIC ALIGNMENT
- ALL FUNCTIONS I²C-BUS CONTROLLED
- ISS FILTER STATUS INFORMATION I²C-BUS READABLE

DESCRIPTION

The TDA 7511 is a high performance tuner circuit for AM/FM car radio. It contains mixers, IF amplifiers, demodulators for AM and FM, quality detection, ISS filter and PLL synthesizer with IF counter on a single chip.

Use of BICMOS technology allows the implementation of several tuning functions and a minimum of external components.

BLOCK DIAGRAM



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PIN DESCRIPTION (conditived)

N°	Pin	Function
1	Z N.W'X1IN2	AM Input2 Mixer1
2	AMMIX1IN1	AM Input1 Mixer1 Reference
	AMRFAGCIN	Input AM RF AGC
4	AMRFAGCOUT	Output AM RF AGC
5	FMPINDR	FM PIN Diode Driver Output
6	FMMOSDR	FM MOS Driver Output
7	FMMIX1IN1	FM Input1 Mixer1
8	GNDRF	RF Ground
9	FMMIX1IN2	FM Input2 Mixer1
10	TV1	Tuning Voltage 1
11	FMRFAGCIN	FM RF AGC Input
12	TV2	Tuning Voltage 2
13	ADJCH	Ident. Adjacent Channel Output
14	FSU	Unweighted Fieldstrength Output

PIN DESCRIPTION (continued)

N°	Pin	Function
15	ISSTC	Time Constant for ISS Filter Switch
16	VCCVCO	VCO Supply
17	GNDVCO	VCO Ground
18	VCOB	VCO Input Base
19	VCOE	VCO Output Emitter
20	DEVTC	Deviation Detector Time Constant
21	XTALD	Xtal Oscillator to MOS Drain
22	XTALG	Xtal Oscillator to MOS Gate
23	GNDVCC3	VCC3 Ground
24	SSTOP	Search Stop Output
25	SDA	I ² C-Bus Data
26	SCL	I ² C-Bus Clock
27	VCC3	Supply Tuning Voltage
28	LPOUT	Op Amp Output to PLL Loop Filters
29	VREF2	Voltage Reference for PLL Op Am
30	LPAM	Op Amp Input to PLL Loo's Filters AM
31	LPFM	Op Amp Input to PLL Loop Filters FM
32	LPHC	High Current PI I. Loop Filter Input
33	GNDVCC1	Digital Grc in 1
34	AMST/MP	AM Sterco Out / Ident. Multipath Output
35	FSW	V /eighted Fieldstrength Output
36	VCC.1	Digital Supply
37	ΜΓΧ/ΑΓΛΜ	MPX Output / AM AF Output
38	∧ MIFREF	Reference Voltage AM IF Amp
39	AMIFBPF	AM IF Filter
25	AMAGC2TC	AM AGC2 Time Constant
41	AMDETC	AM Detector Capacitor
42	MUTETC	Softmute Time Constant
43	AMIF2IN	Input AM IF2
44	FMDEMC	FM Demodulator Reference
45	FMMIX2IN2	FM IF1 MIX2 Input1
46	FMMIX2IN1	FM IF1 MIX2 Input2
47	GNDDEM	Ground FM Demodulator
48	VREF1	Reference 5V
49	GNDVCC2	Analog Ground
50	FMAMP2OUT	FM IF1 Amplifier2 Output



N°	Pin	Function
51	VCC2	Analog Supply
52	FMAMP2IN	FM IF1 Amplifier2 Input
53	FMIF1REF	FM IF1 Amplifier Reference
54	FMAMP1OUT	FM IF1 Amplifier1 Output
55	AMMIX2OUT2	AM Tank 450kHz
56	AMMIX2OUT1	AM Tank 450kHz
57	FMAMP1IN	FM IF1 Amplifier1 Input
58	AMIF1IN	AM IF1 Input
59	GNDIF1AMP	FM IF1 Amplifier Ground
60	FMIF1AGCIN	FM IF1 AGC Input
61	MIX1OUT2	MIX Tank 10.7MHz
62	MIX1OUT1	MIX Tank 10.7MHz
63	AMRFAGCTC	AM RF AGC Time Constant
64	AMPINDR	AM PIN Diode Driver Output
SOLU	ITE MAXIMUM RA	ATINGS

PIN DESCRIPTION (continued)

ABSOLUTE MAXIMUM RATINGS

Symbol	Parameter		Value	Unit
Vs	Supply Voltage		10.5	V
T _{amb}	Ambient Temperature	abs	-40 to 85	°C
T _{stg}	Storage Temperature	0	-55 to +150	°C

THERMAL DATA

	*(5)		
Symbol	Parameter	Value	Unit
R _{Th(j-an-b)} Fihermal resist	ance	68	°C/W

ELECTRICAL CHARACTERISTCS

 $T_{amb} = +25^{\circ}C, V_{CC1} = V_{CC2} = V_{CC3} = V_{CCVCO} = V_{CCMIX1} = 8.5V, f_{RF} = 98MHz, dev. = 40kHz, f_{MOD} = 1kHz, f_{IF1} = 10.7MHz, f_{IF2} = 450KHz, f_{Xtal} = 10.25MHz, in test or application circuit, unless otherwise$ specified.

Symbol	Parameter	Test Condition	Min.	Тур.	Max.	Unit
Supply						
V _{CC1}	Digital supply voltage		7.5	8.5	10	V
V _{CC2}	Analog supply voltage		7.5	8.5	10	V
V _{CC3}	Analog tuning voltage		7.5	8.5	10	V
Vccvco	VCO supply voltage		7.5	8.5	10	V

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Symbol	Parameter	Test Condition	Min.	Тур.	Max.	Unit
V _{CCMIX1}	MIX1 supply voltage		7.5	8.5	10	V
V _{CCMIX2}	MIX2 supply voltage		7.5	8.5	10	V
I _{CC1}	Supply current	FM ON		7.5		mA
I _{CC1}	Supply current	AM ON		10		mA
I _{CC2}	Supply current	FM ON		50		mA
I _{CC2}	Supply current	AM ON		60	10	r.ıA
I _{CC3}	Supply current			2	777	mA
Iccvco	Supply current			2.5		mA
ICCMIX1	Supply current	FM ON	~	6		mA
I _{CCMIX1}	Supply current	AM ON		5	1	mA
I _{CCMIX2}	Supply current	AM ON		5	6	mA
Referenc	e Voltages	<u></u>		0,		
V _{REF1}	Internal reference voltage	I _{REF1} = 0mA	0	5		V
V _{REF2}	Internal reference voltage	I _{REF2} = 0 nA		2.5		V
Wide Bar	nd RF AGC					
V ₁₁	Threshold AGC start	$V_3 = V_{CC2}/2$		80		dBµ
R _{IN}	Input resistance			500		Ω
C _{IN}	Input capacitance	0,		2.5		pF
Narrow B	and RF & Kevir g AC C			L		
V ₆₀	Lower thre: nold	FMAGC, V ₁₁ = 0mV _{RMS}		82		dBµ
V ₆₀	Up, e. th reshold	FMAGC, V ₁₁ = 0mV _{RMS}		92		dBµ
RIN	Input resistance			10		kΩ
	Input capacitance			2.5		pF
	SFET Driver Output	1	I	I	I	1
V ₆	Max. AGC output voltage	V ₁₁ = 0mV _{RMS}	V _{CC2} -0.5V			V
V ₆	Min. AGC output voltage	$V_{11} = 50 m V_{RMS}$			0.5	V
I ₆	Min. AGC charge current	$V_{11} = 0mV_{RMS}, V_6 = V_{CC2}/2$		-12.5		μA
I ₆	Max. AGC discharge current	$V_{11} = 50 \text{mV}_{\text{RMS}}, V_6 = V_{\text{CC2}}/2$		1.25		mA
AGC PIN	Diode Driver Output	1	I	I	I	1
I ₅	AGC OUT, current min.	$V_{11} = 0mV_{RMS}, V_5 = 2V$		50		μA
I ₅	AGC OUT, current max.	$V_{11} = 50 \text{ mV}_{RMS}, V_5 = 2V$		-6		mA
Mixer1 (1	0.7MHz)	<u>I</u>	L	1		1
	Input impedance	Balanced, f = 98MHz				Ω

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Symbol	Parameter	Test Condition	Min.	Тур.	Max.	Unit
IP3	3rd order intercept point			122		dBµV
F	Noise figure			6		dB
A _C	Conversion gain			120		mS
IF1 Ampl	ifier1 & 2 (10.7MHz)					
A _{min}	Min. gain	IFG		9		dB
A _{max}	Max. gain	IFG		15	.10	ι'В
R _{IN}	Input resistance			330	775	Ω
R _{OUT}	Output resistance			3.0		Ω
P _{1dB}	1dB compression point	Output referred		120		dBµV
IP3	3rd order Intercept Point	Output referred		132	- 2 (9	dBµ∨
Mixer2 (4	50kHz)	-*8	<u> </u>		C	
R _{IN}	Input impedance	101		330		Ω
V ₄₆	Max. input voltage	~~~~~~~~~~~~~~~~~~~~~~~~~~~~~~~~~~~~~~	25	900		mV _{RN} S
V ₄₈	Limiting sensitivity	S/N = 20cB		25		μV
А	Mixer gain	10		18		dB
Limiter 1	(450kHz)	51 . 60'		I	I	
G _{Limiter}	Gain			80		dB
Demodul	ator, Audio Output			I	I	
THD	0	Dev.= 75kHz, V ₄₆ = 10mV _{RMS}			0.1	%
V _{MPX}	MPX output signal	Dev.= 75kHz		500		mV _{RN} S
ROUT	Ou put resistance			350		Ω
	DC offset fine adjust	DEM, MENA=1		6		mV
!∆V _{max}	DC offset fine adjust	DEM, MENA=1		186		mV
S/N		Dev.= 40kHz,V ₄₆ = 10mV _{RMS}		75		dB
Quality D	etection					
S-meter,	Unweighted Fieldstrength					
V46	Min. input voltage MIX2			10		μV
ΔV	Per decade	SMSL = 0		1		V
ΔV	Per decade	SMSL = 1		1.5		V
V ₁₄	Fieldstrength output	$V_{46} = 0V_{RMS}$		0.1		V
V ₁₄	Fieldstrength output	$V_{46} = 1V_{RMS}$		4.9		V
R _{OUT}	Output resistance			4		kΩ

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TK Temp coeff. 0 V35 Fieldstrength output V46 = 0VRMS 2.5 V35 Fieldstrength output V46 = 1VRMS 4.9 ROUT Output resistance 12 12 Adjacent Channel Gain 4.9 52 1 Amin Gain minimum ACG 52 1 Amax Gain maximum ACG 52 1 Amax Gain maximum ACG 52 1 Amax Gain maximum ACG 52 1 Adjacent Channel Filter 100 52 1 fHP -3dB frequency highpass ACF 100 1 f-20dB Attenuation 20dB 70 1 2 Adjacent Channel Output 12 1 1 1 V13 Output voltage high 1 1 1 ROUT Output voltage high 1 1 1 Amin Gain minimum MPG 12 1	ti	ion	Min.	Тур.	Max.	Unit
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V ₃₄ Output voltage low 0.1 V ₃₄ Output voltage high 4.9 R _{OUT} Output resistance 2.5 ISS (intelligent Selectivity System) Filter 450kHz			5		10	
V34 Output voltage high 4.9 R _{OUT} Output resistance 2.5 ISS (intelligent Selectivity System) Filter 450kHz						
R _{OUT} Output resistance 2.5 ISS (intelligent Selectivity System) Filter 450kHz				0.1		V
ISS (intelligent Selectivity System) Filter 450kHz				4.9		V
Filter 450kHz				2.5		kΩ
· · · · · · · · · · · · · · · · · · ·					I	
f _{centre} Centre frequency f _{REF intern} = 450kHz 450						
				450		kHz
BW 3dB Bandwidth, -3dB ISS80 = 1 80				80		kHz
BW Bandwidth, -20dB ISS80 = 1 150 20dB 150 150 150				150		kHz

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Symbol	Parameter	Test Condition	Min.	Тур.	Max.	Uni
BW 3dB	Bandwidth, -3dB	ISS80 = 0		120		kHz
BW 20dB	Bandwidth, -20dB	ISS80 = 0		250		kHz
BW 3dB	Bandwidth weather band	ISS30 = 1		30		kHz
BW 20dB	-20dB weather band	ISS30 = 1		80		kHz
Δf_{min}	Fine adjust	AISS		-20	.19	5 k Hz
Δf_{max}	Fine adjust	AISS		10	672	kH:
Adjacent	Channel ISS Filter Threshold				<u> </u>	1
V _{NTH}	Internal low threshold	ACNTH	5	00		V
V _{NTH}	Internal high threshold	ACNTH		0.3		ν v
Vwth	Internal low threshold	ACWTH		0.25	0	V
V _{WTH}	Internal high threshold	ACWTH		0.95		V
Multipat	h Threshold	-5	0			
VTHMP	Internal low threshold	MPTH		0.50		V
V _{THMP}	Internal high threshold	МРТН		1.25		V
ISS Filte	r Time Constant	5 00				
I ₁₅	Charge current low mid	TISS, ISSCTL = 1		-74		μA
I ₁₅	Charge current high min	TISS, ISSCTL = 1		-60		μA
I ₁₅	Charge current k w harrow	TISS, ISSCTL = 1		-124		μA
I ₁₅	Charge cui ent high narrow	TISS, ISSCTL = 1		-110		μA
I ₁₅	Dis.th.trge current low	TISS, ISSCTL = 0		1		μA
I ₁₅	Lischarge current high	TISS, ISSCTL = 0		15		μA
	Low voltage	ISSCTL = 0		0.1		V
V ₁₅	High voltage	ISSCTL = 1		4.9		V
ISS Filter	Switch Threshold		1	1	1	1
V ₁₅	Threshold ISS on	ISSCTL = 0		3		V
V ₁₅	Threshold ISS off	ISSCTL = 0		1		V
V ₁₅	Threshold ISS narrow on	ISSCTL = 0		4		V
V ₁₅	Threshold ISS narrow off	ISSCTL = 0		2		V
I ₂₀	Charge current low	TDEV		-20		μA
I ₂₀	Charge current high	TDEV		-34		μA
I ₂₀	Discharge current low	TDEV		6		μA
	Discharge current high	TDEV				<u> </u>

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Symbol	Parameter	Test Condition	Min.	Тур.	Max.	Unit
DEV _{WTH}	Internal low threshold	DWTH		30		kHz
DEV _{WTH}	Internal high threshold	DWTH		75		kHz
RATIO _{mi}	Referred to threshold	DTH		1		
RATIO _m	Referred to threshold	DTH		1.5		
Softmute						5
Vant	Upper startpoint	SMTH, SMD, SLOPE = 0		16	61	dBµV
V _{ANT}	lower startpoint	SMTH, SMD, SLOPE = 0				dBµV
asMmin	Min. softmute depth	SMD, SLOPE = 0, SMTH _{Upper}	~	18		dB
asMmax	Max. softmute depth	SMD, SLOPE = 0, SMTH _{Upper}		36	12	dB
a _{SMTHIS} S	Mute depth threshold for ISS filter on	SMCTH	0.2	2	2	dB
V _{ACTH}	Internal AC mute threshold	ACM	60	0	220	mV
asmac	AC mute depth	000	Y	6		dB
I ₄₂	Charge current	<u> </u>		-47.5		μA
I ₄₂	Discharge current			2.5		μA
S/N Over	All	5				4
S/N	- codulo.	V _{IN_min} = 60dBµV, dev.= 40kHz,LP=15KHz deemphasis t = 50µs		66		dB

ELECTRICAL CHARACTERISTICS

 $T_{amb} = +25^{\circ}\text{G}, \quad V_{CC1} = V_{CC2} = V_{CC3} = V_{CCVCO} = V_{CCMIX1} = V_{CCMIX2} = 8.5\text{V}, \quad f_{RF} = 1\text{MHz}, \quad f_{MOD} = 400\text{Hz} \text{ at } 30\% \text{ AMf}_{IF1} = 1\text{C} II\text{Hz}, \quad f_{IF2} = 450\text{kHz}, \quad f_{xtal} = 10.25\text{MHz}, \quad \text{in test or application circuit, (unless otherwise noted, V_{inRF} antenr a input).}$

οσιπίζ	Parameter	Test Condition	Min.	Тур.	Max.	Unit
AM SECT	ION					
Global	10					
VINRF	Max. sensitivity	$\begin{array}{l} \text{Ref.: } V_{\text{INRF}} = 74 \text{dB}\mu\text{V}, \\ \Delta\text{V}_{37} = -10 \text{dB} \end{array}$		19		dBμV
V _{INRF} us	Usable sensitivity	(S+N)/N = 20 dB	30	26		dBµV
ΔVINRF	AGC Range	$\begin{array}{l} \text{Ref.: } V_{\text{INRF}} = 74 \text{dB}\mu\text{V}, \\ \Delta\text{V}_{37} = -10 \text{dB} \end{array}$	51			dB
(S+N)/N	Signal to Noise Ratio	Ref.: V _{INRF} = 74dBµV	49	56		dB

 $T_{amb} = +25^{\circ}C, V_{CC1} = V_{CC2} = V_{CC3} = V_{CCVICO} = V_{CCMIX1} = V_{CCMIX2} = 8.5V, f_{RF} = 1MHz, f_{MOD} = 400Hz at 30\%$ AMf_{IF1} = 10.7MHz, f_{IF2} = 450kHz, f_{xtal} = 10.25MHz, in test or application circuit, (unless otherwise noted, V_{inRF} antenna input).

Symbol	Parameter	Test Condition	Min.	Тур.	Max.	Unit
a _{lF}	IF rejection	Ref: $V_{INRF} = 74dB\mu V$, IF1 = 10.7MHz IF2 = 450kHz $\Delta V_{37} = -10dB$	100 100			dB dB
a _{TW}	Tweet	C ₂₈ = 22µF	-3	-0.7		dB
f _{AF}	Frequency response	Ref.: $V_{INRF} = 74 dB\mu V$, $\Delta V_{AF} = -3 dB$		3.6		kHz
VINRFSS	Seek Stop Sensitivity	AMSS	20	35	15	dBµV
THD	Total Harmonic Distortion	$V_{INRF} = 74 \text{ dB}\mu\text{V}, m = 0.8$ m = 0.3 V_{INRF} = 120 db $\mu\text{V}, m = 0.8$ m = 0.3	Pr	0.5 0.3 1.0 0.3	d	%
V ₃₇	Output level	V _{INRF} = 74 dBµV		220		mV _{RM} S
V ₃₄	Output level	V _{INRF} = 74 dBµ n	61	190		mV _{RM} S
V ₃	Min. RF AGC threshold Max. RF AGC threshold	AMAGC Wide	1	82 98.8		dBμV dBμV
V ₅₈	Min. IF AGC threshold Max. IF AGC threshold	A'.MA.GC Middle		80.1 96.8		dΒμV dBμV
V ₃	Min. IF AGC thresho'a Max. IF AGC threshold	DAGC Narrow		32.6 80.7		dBμV dBμV
R _{63OUT}	Output impodance	51		100		kΩ
R _{40OUT}	Ouipurinpedance	AMSEEK = 0		150		kΩ
R ₄₀₀ JT	Output impedance	AMSEEK = 1		5		kΩ
AGC Vol	tage Driver Output					
V ₄	Max. AGC output voltage		3.5			V
V ₄	Min. AGC output voltage				0.5	V
14	AGC current			100		μΑ
AGC PIN	Diode Driver Output					l
I ₆₄	AGC driver current			1.5		mA
AM Mixer	1 (10.7MHz)	1	1		1	1
R _{IN}	Input impedance			1.2		kΩ
IP3	3rd order intercept point			140		dBµV
F	Noise figure			7		dB

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 $T_{amb} = +25^{\circ}C, V_{CC1} = V_{CC2} = V_{CC3} = V_{CCVCO} = V_{CCMIX1} = V_{CCMIX2} = 8.5V, f_{RF} = 1MHz, f_{MOD} = 400Hz at 30\%$ AMf_{IF1} = 10.7MHz, f_{IF2} = 450kHz, f_{xtal} = 10.25MHz, in test or application circuit, (unless otherwise noted, V_{inRF}) antenna input).

Symbol	Parameter	Test Condition	Min.	Тур.	Max.	Unit
А	Gain			6		dB
C _{min}	Min. capacitance step	IF1T		0.55		pF
C _{max}	Max. capacitance	IF1T		8.25		pF
C ₆₁₋₆₂		IF1T		2		pF
AM Mixer	2 (450kHz)				10	5
R _{IN}	Input impedance	Dependent on application		5	775	kΩ
IP3	3rd order intercept point			1-17		dBµV
F	Noise figure		20	12	10	dB
А	Max. gain	Mixer2 tank output		15	15	dB
ΔΑ	Gain control range	1010		20	X	dB
C _{min}	Min. cap step	IF2T	5	1.6		pF
C _{max}	Мах. сар	IF2T		24		pF
C ₅₅₋₅₆		IF2T	1	2		pF
ELECTRI	CAL CHARACTERISTICS	SI absolu	I	1	I	
0	Demonstration of the second seco	Test Condition		T		11

Symbol	Paramete.	Test Condition	Min.	Тур.	Max.	Unit
ADDITIO	NAL PARAMETFF S					1
Outputs	of Tuning Vູ ທີ່ເສຍ(TV1, TV2)	51				
Vout	Oupurvoltage	TVR,TVO	1		VCC3- 1V	V
Four	Output impedance			20		kΩ
Xtal Refe	rence Oscillator	1		1		1
fLO	Reference frequency	C _{Load} = 15pF		10.25		MHz
CStep	Min. cap step	XTAL		0.6		pF
C _{max}	Мах. сар	XTAL		19.4		pF
$\Delta f/f$	Freq. deviation versus VCC2	$\Delta V_{CC2} = 1 V$		1.5		ppm/\
$\Delta f/f$	Freq. deviation versus temp	-40°C < T < +85°C		0.2		ppm/k
I ² C-Bus i	nterface	1				
fSCL	Clock frequency				400	kHz
VIL	Input low voltage				1	V

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l _{IN} In		Test Condition	Min.	Тур.	Max.	Unit
	nput high voltage		3			V
V _O O	nput current		-5		5	μA
	Output voltage SDA acknowledge	I _O = 1.6mA			0.4	V
oop Filter	Input/Output			I		
-l _{IN} In	nput leakage current	V _{IN} = GND, PD _{OUT} = Tristate	-0.1		0.1	μA
l _{IN} In	nput leakage current	V _{IN} = VREF1 PD _{OUT} = Tristate	-0.1		0.1	μΑ
V _{OL} O	Output voltage Low	I _{OUT} = -0.2mA		0.05	<u>ر</u> ک	Γ v
V _{OH} O	Dutput voltage High	I _{OUT} = 0.2mA	VCC3 -0.5	VC.C.3 - 1.0.7		V
I _{OUT} O	Output current, sink	$V_{OUT} = 1V$ to V_{CC3} -1V	2/		10	mA
I _{OUT} O	Output current, source	$V_{OUT} = 1V$ to V_{CC3} -1V	-10		CV)	mA
/oltage Co	ntrolled Oscillator (VCO)	00		0,		•
VCOmin M	Iinimum VCO frequency	205	50			MH
VCOmax M	laximum VCO frequency	0 * 20			200	MH
C/N C	Carrier to Noise	1l'Hz offset		85		dBo
STOP Out	put	5 50				
V ₂₄ 0	Output voltage low	I ₂₄ = -20μA			0.2	V
V ₂₄ 0	Dutput voltage hign	I ₂₄ = 20μA	3			V
V ₄₆ IF	F counter Sensitivity	Antenna input		6		dBµ

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1 FUNCTIONAL DESCRIPTION

1.1 FM Section

1.2 Mixer1, AGC and 1.IF

Mixer1 is a wide dynamic range stage with low noise and large input signal performance. The mixer1 tank can be adjusted by software (IF1T). The AGC operates on different sensitivities and bandwidths (FMAGC) in order to improve the input sensitivity and dynamic range (keying AGC). The output signals of AGC are controlled voltage and current for preamplifier and prestage pin diode attenuator. (look at Figure 4)

Two 10.7MHz programmable amplifiers (IFG1, IFG2) correct the IF ceramic insertion loss and the costumer level plan application.

1.3 Mixer2, Limiter and Demodulator

In this 2. mixer stage the first 10.7MHz IF is converted into the second 450kHz IF. A multi-stage limiter generates signals for the complete integrated demodulator without external tank. MPX output DC offset compensation is possible by software (DEM).

1.4 Quality Detection and ISS (look at Figure 2)

Fieldstrength

Parallel to mixer2 input a 10.7MHz limiter generates a signal for digital IF counter and a fieldstrength output signal. This internal unweighted fieldstrength is used for keying AGC, cdiacent channel and multipath detection and is available at PIN14 (FSU) after +6dB buffer stage. The behaviour of this output signal can be corrected for DC offset (SL) and slope (SMSL). The internal generated unweighted fieldstrength is filtered at PIN35 and used for softmute function and generation of ISS filter switching signal for weak input level (sm).

Adjacent Channel Detector

The input of the adjacent channel detector is AC coupled from internal unweighted fieldstrength. A programmable highpass or bandpass (ACF) and amplifier (ACG) as well as rectifier determines the influences. This voltage is compared with adjustable comparator1 thresholds (ACWTH, ACNTH). The output signal of this comparator generates a DC level at PIN15 by programmable time constant. Time control (TISS) for a present adjacent channel is made by charge and discharge current after comparator1 in an external capacitance. The charge current is fixed and the discharge current is controlled by I²C Bus. This level produces digital signals (ac, ac+) in an additional comparator1. The adjacent channel information is available as analog output signal after rectifier and +8dB output b.ther.

Multir at' Detector

Tige input of the multipath detector is AC coupled from internal unweighted fieldstrength. A programmable bandpass (MPF) and amplifier (MPG) as well as rectifier determines the influences. This voltage is compared with an adjustable comparator2 thresholds (MPTH). The output signal of this comparator2 is used for the "Milano" effect. In this case the adjacent channel detection is switched off. The "Milano" effect is selectable by I²C Bus (MPOFF). The multipath information is available as analog output signal after rectifier and +8dB output buffer.

450kHz IF Narrow Bandpass Filter (ISS filter)

The device gets an additional second IF narrow bandpass filter for suppression noise and adjacent channel signal influences. This narrow filter has three switchable bandwidthes, narrow range of 80kHz, mid range of 120kHz and 30KHz for weather band information. Without ISS filter the IF bandwidth (wide range) is defined only by ceramic filter chain. The filter is switched in after mixer2 before 450kHz limiter stage. The centre frequency and matching to the demodulator center frequency can be fine adjusted (AISS) by software..

Deviation Detector

In order to avoid distortion in audio output signal the narrow ISS filter is switched OFF for present overdeviation.



Hence the demodulator output signal is detected. A lowpass filtering and peak rectifier generates a signal that is defined by software controlled current (TDEV) in an external capacitance. This value is compared with a programmable comparator3 thresholds (DWTH, DTH) and generates two digital signals (dev, dev+).

ISS Switch Logic

All digital signals coming from adjacent channel detector, deviation detector and softmute are acting via switching matrix on ISS filter switch. The IF bandpass switch mode is controlled by software (ISSON, ISS30, ISS80, ISSCTL). The switch ON of the IF bandpass is also available by manipulation of the voltage at PIN15. Two application modes are available (APPM). The conditions are described in table 37.

1.5 Soft Mute Control

The external fieldstrength signal at PIN 35 is the reference for mute control. The startpoint, mute depth and slope are programmable (SMTH, SMD, SLOPE) in a wide range. The time constant is defined by external capacitance. Additional adjacent channel mute function is supported. A highpass filter with -3dB threshold flequency of 100kHz, amplifier and peak rectifier generates an adjacent noise signal from MPX output with the same time constant for softmute. This value is compared with comparator5 thresholds (ACM). For present strong adjacent channel the MPX signal is attenuated typical 6dB.

1.6 AM Section

The upconversion mixer1 is combined with a gain control circuit 1 sensing three input signals, narrow band information at PIN 39, upconversion signal at PIN 58 and wide band information at PIN 3. This gain control circuit gives two output signals. The first one is a current for pin diode attenuator and the second one is a voltage for preamplifier. It is possible to put in a separate narrow bandpess ulter before mixer2 at PIN 58. The intervention point for first AGC (AMAGC) is programmable by software.

The oscillator frequency for mixer1 is generated by dividing the FM VCO frequency (AMD).

In mixer2 the IF1 is downconverted into the IF2 450kHz. Before the output signal reaches the 450kHz tank an attenuator for IF gain control 2 is passed Mixer1 and mixer2 tanks are software controlled adjustable (IF1T, IF2T).

After filtering by ceramic filter a 450 k 47 amplifier with a gain control 3 is included. The gain control 2 and 3 are the second AGC and programmable too by software (DAGC). In order to avoid an oscillation in intervention point it is important to know that the DAGC threshold has to be smaller than AMAGC!

The demodulator is a peak detector. A further time constant with capacitor at pin40 produces a DC AGC reference voltage dependent on input signal. The time constant is switchable by ratio of 30. This is necessary for the station search function. The switching is software controlled (AMSEEK).

An internal comparator compares the AGC voltage with a programmable reference (AMSS). Consequently it is possible to generate a seekstop impulse over a defined range.

A separate output is available for AMIF stereo or a permanent seek stop signal(SSTSEL).

1.7 PLL and IF Counter Section

PLL Frequency Synthesizer Block

This part contains a frequency synthesizer and a loop filter for the radio tuning system. Only one VCO is required to build a complete PLL system for FM and AM upconversion. For auto search stop operation an IF counter system is available.

The counter works in a two stages configuration. The first stage is a swallow counter with a two modulus (32/33) precounter. The second stage is an 11-bit programmable counter.

The circuit receives the scaling factors for the programmable counters and the values of the reference frequencies via an I²C-Bus interface. The reference frequency is generated by an adjustable internal (XTAL) oscillator followed by the reference divider. The reference and step-frequencies are free selectable (RC, PC).

Output signals of the phase detector are switching the programmable current sources. The loop filter integrates



their currents to a DC voltage.

The values of the current sources are programmable by 6 bits also received via the I²C Bus (A, B, CURRH, LPF). To minimize the noise induced by the digital part of the system, a special guard area is implemented.

The loop gain can be set for different conditions by setting the current values of the chargepump generator.

Frequency Generation for Phase Comparison

The RF signals applies a two modulus counter (32/33) pre-scaler, which is controlled by a 5-bit divider(A). The 5-bit register (PC0 to PC4) controls this divider. In parallel the output of the prescaler connects to an 11-bit divider(B). The 11-bit PC register (PC5 to PC15) controls this divider

Dividing range: $f_{OSC} = (R+1) \times f_{REF}$ $f_{VCO} = [33 \times A + (B + 1 - A) \times 32] \times f_{REF}$ $f_{VCO} = (32 \times B + A + 32) \times f_{REF}$ Important: For correct operation: $A \le 32$; $B \ge A$

Three State Phase Comparator

The phase comparator generates a phase error signal according to phase dimensions between f_{SYN} and f_{REF} . This phase error signal drives the charge pump current generator.

Charge Pump Current Generator

This system generators signed pulses of current. The phase or or signal decides the duration and polarity of those pulses. The current absolute values are programmable by A register for high current and B register for low current.

Inlock Detector

Switching the chargepump in low current niccle can be done either via software or automatically by the inlock detector, by setting bit LDENA to "1"

After reaching a phase difference of 10 - 40nsec and a delay of some times 1/f_{REF}, the chargepump is forced in low current mode. A new P L divider alternation by I²C-Bus will switch the chargepump in the high current mode.

Few programmable µ'ase errors (D0, D1) are available for inlock detection

The count of cotocted inlock informations, to release the inlock signal is adjustable (D2, D3), to avoid a switching to low current ouring a frequency jump.

Low Noise CMOS Op-amp

In internal voltage divider at pin VREF2 connects the positive input of the low noise op-amp. The charge pump output connects the negative input. This internal amplifier in cooperation with external components can provide an active filter. The negative input is switchable to three input pins, to increase the flexibility in application. This feature allows two separate active filters for different applications.

A logical "0" in the LPF register activates PIN LPFM, otherwise PIN LPAM is active. While the high current mode is activated LPHC is switched on.

IF Counter Block

The input signal for FM and AM upconversion is the same 10.7MHz IF level after limiter. The grade of integration is adjustable by eight different measuring cycle times. The tolerance of the accepted count value is adjustable, to reach an optimum compromise for search speed and precision of the evaluation.

For the FM range the center frequency of the measured count value is adjustable in 32 steps, to get the possibility of fitting the IF-filter tolerance. In the AM upconversion range an IF frequency of 10.689MHz to 10.720MHz with 1kHz steps is available.

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The IF-Counter Mode

The IF counter works in 2 modes controlled by IFCM register.

Sampling Timer

A sampling timer to generate the gate signal for the main counter is build with a 14-bit programmable counter (IRC). In FM mode a 6.25kHz, in AM mode a 1kHz basically signal is generated. This is followed by an asynchronous divider to generate several sampling times.

Intermediate Frequency Main Counter

This counter is a 11 - 21-bit synchronous autoreload down counter. Five bits (CF) are programmable to have the possibility for an adjust to the frequency of the IF-filter. The counter length is automatic adjusted to the chosen sampling time and the counter mode (FM, AM-UPC).

At the start the counter will be loaded with a defined value which is an equivalent to the divider value (t_{Sample} x f_{IF}).

If a correct frequency is applied to the IF counter frequency input at the end of the sampling time the main counter is changing its state from 0h to 1FFFFh.

-OW te This is detected by a control logic and an external search stop output is changing from LOW to HIGH. The frequency range inside which a successful count result is adjustable by the EW Lits.

$t_{TIM} = (IRC + 1) / f_{OSC}$

 $t_{CNT} = (CF + 1697) / f_{IF}$ FM mode

 $t_{CNT} = (CF + 10689) / f_{IF}$ AM up conversion mode

Counter result succeeded:

t_{TIM} ≥ t_{CNT} - t_{ERR}

 $t_{TIM} \leq t_{CNT} + t_{ERF}$

Counter result failed:

T. M > CNT + tERR

ttim < tcnt - tfrr

t_{TIM} = IF timer cycle time

t_{CNT} = IF counter cycle time

t_{ERR} = discrimination window (controlled by the EW registers)

The IF counter is only started by inlock information from the PLL part. It is enabled by software (IFENA).

Adjustment of the Measurement Sequence Time

The precision of the measurements is adjustable by controlling the discrimination window. This is adjustable by programming the control registers EW0 to EW2.

The measurement time per cycle is adjustable by setting the Register IFS0 - IFS2.

Adjust of the Frequency Value

The center frequency of the discrimination window is adjustable by the control register CF0 to CF4.



1.8 I²C-Bus Interface

The TDA 7511 supports the I²C-Bus protocol. This protocol defines any device that sends data onto the bus as a transmitter, and the receiving device as the receiver. The device that controls the transfer is a master and device being controlled is the slave. The master will always initiate data transfer and provide the clock to transmit or receive operations.

Data Transition

Data transition on the SDA line must only occur when the clock SCL is LOW. SDA transitions while SCL is HIGH will be interpreted as START or STOP condition.

Start Condition

A start condition is defined by a HIGH to LOW transition of the SDA line while SCL is at a stable HIGH level. This "START" condition must precede any command and initiate a data transfer onto the bus. The "DA 7511 continuously monitors the SDA and SCL lines for a valid START and will not response to any command it this condition has not been met.

Stop Condition

A STOP condition is defined by a LOW to HIGH transition of the SDA while the SCL line is at a stable HIGH level. This condition terminates the communication between the devices and forces the bus-interface of the TDA 7511 into the initial condition.

Acknowledge

Indicates a successful data transfer. The transmitter will release the bus after sending 8 bits of data. During the 9th clock cycle the receiver will pull the SDA line to LOW level to indicate it receive the eight bits of data.

Data Transfer

During data transfer the TDA 7511 samples the SDA line on the leading edge of the SCL clock. Therefore, for proper device operation the SDA line π_{1} us the stable during the SCL LOW to HIGH transition.

Device Addressing

To start the communication between two devices, the bus master must initiate a start instruction sequence, followed by an eight bi word corresponding to the address of the device it is addressing.

The most significant 6 bits of the slave address are the device type identifier.

The TDA 7511 device type is fixed as "110001".

The next significant bit is used to address a particular device of the previous defined type connected to the bus.

The state of the hardwired PIN 41 defines the state of this address bit. So up to two devices could be connected on the same bus. When PIN 41 is connected to VCC2 the address bit "1" is selected. In this case the AM part doesn't work. Otherwise the address bit "0" is selected (FM and AM is working). Therefor a double FM tuner concept is possible.

The last bit of the start instruction defines the type of operation to be performed:

- When set to "1", a read operation is selected

- When set to "0", a write operation is selected

The TDA 7511 connected to the bus will compare their own hardwired address with the slave address being transmitted, after detecting a START condition. After this comparison, the TDA 7511 will generate an "acknowledge" on the SDA line and will do either a read or a write operation according to the state of R/W bit.

Write Operation

Following a START condition the master sends a slave address word with the R/W bit set to "0". The TDA 7511 will generate an "acknowledge" after this first transmission and will wait for a second word (the word address field). This 8-bit address field provides an access to any of the 32 internal addresses. Upon receipt of the word address the TDA 7511 slave device will respond with an "acknowledge". At this time, all the following words transmitted to the TDA 7511 will be considered as Data. The internal address will be automatically incremented. After each word receipt the TDA 7511 will answer with an "acknowledge".

Read Operation

IF the master sends a slave address word with the R/W bit set to "1", the TDA 7511 will transit one 8-bit data word. This data word includes the following informations:

bit0 (ISS filter, 1 = ON, 0 = OFF)

bit1 (ISS filter bandwidth, 1 = 80kHz, 0 = 120kHz)

bit2 (MPOUT,1 = multipath present, 0 = no multipath)

bit3 (1 = PLL is locked in , 0 = PLL is locked out).

bit4 (fieldstrength indicator, 1 = lower as softmute threshold, 0 = higher as softmute threshold)

Josur Product bit5 (adjacent channel indicator, 1 = adjacent channel present, 0 = no adjacer c ciannel)

bit6 (deviation indicator, 1 = strong overdeviation present, 0 = no strong overdeviation)

bit7 (deviation indicator, 1 = overdeviation present, 0 = no overdeviation).

Software Specification 2

The interface protocol comprises:

- start condition (S)
- chip address byte
- subaddress byte
- sequence of data (N bytes + Acknowledge)

- stop condition (P) Figure 1. CHIP ADDRESS SUBADDRESS DATA1 - DATAn MSB LSB MSB LSB MSB LSB s 1 0 0 0 Х ACK Т Т A0 ACK DATA ACK Ρ 1 0 1 A3 A1 ACK = Acknowledge = Start S P = Stop = Pagemode = R/W Bit Х 99AT0054 19/41

2.1 Address Organization

Table 1.

Function	Addr	7	6	5	4	3	2	1	0
CHARGEPU MP	0	LPF	CURRH	B1	B0	A3	A2	A1	A0
LOCKDET	1	LDENA	D3	D2	D1	D0	AMON	TEST3	RES2
PLL	2	PC7	PC6	PC5	PC4	PC3	PC2	PC1	PC0
COUNTER	3	PC15	PC14	PC13	PC12	PC11	PC10	PC9	PC8
PLL REF	4	RC7	RC6	RC5	RC4	RC3	RC2	RC1	RC0
COUNTER	5	RC15	RC14	RC13	RC12	RC11	RC10	RC9	RC8
TV1	6	TV011	TV010	TVR15	TVR14	TVR13	TVR12	TVR11	TV.R10
TV2	7	TV021	TV020	TVR25	TVR24	TVR23	TVR22	TVR21	TV/R20
IFC CTRL 1	8	TV013	TV012	TV023	TV022	IFENA	EW2	FV '1	EW0
IFC CTRL 2	9	IFS2	IFS1	IFS0	CF4	CF3	CF2	CP1	CF0
IF REF CNT1	10	IRC7	IRC6	IRC5	IRC4	IRC3	IRC	IRC1	IRC0
IF REF CNT2	11	IFCM1	IFCM0	IRC13	IRC12	IRC11	112010	IRC9	IRC8
IF1/FMAGC	12	-	FMAGC2	FMAGC1	FMAGC0	IFG21	JFG20	IFG11	IFG10
DEM ADJ	13	DNB1	DNB0	DEM5	DEM4	DE1.13	DEM2	DEM1	DEM0
QUALITY AC	14	ACNTH1	ACNTH0	ACWTH2	ACWTH1	лс∵∕тно	ACG	ACF	ISS30
QUALITY MP	15	MPAC	APPM2	APPM1	MPT H1	MPTH0	MPG	MPF	MPOFF
QUALITYDEV	16	-	DTH1	DTH0	CW.TH1	DWTH0	TDEV2	TDEV1	TDEV0
QUALITYISS	17	AISS1	AISS0	TISS2	TISS1	TISS0	ISS80	ISSON	ISSCTL
AM CTL1	18	DAGC3	DAGC2	DAGC1	DAGC0	AMD1	AMD0	AMST	AMSEEK
AM CTL2	19	AMSS3	AMSS2	AI ISS1	AMSS0	AMAGC3	AMAGC2	AMAGC1	AMAGC0
MUTE1	20	SMCTH1	SMICT.HO	SLOPE	MENA	SMD3	SMD2	SMD1	SMD0
MUTE2	21	ACM3	.\CM2	ACM1	ACM0	SMTH3	SMTH2	SMTH1	SMTH0
SLIDER	22		-	SL5	SL4	SL3	SL2	SL1	SL0
TANK ADJ	23	1-1-13	IF1T2	IF1T1	IF1T0	IF2T3	IF2T2	IF2T1	IF2T0
XTAL ADJ	24	-		CLKSEP	XTAL4	XTAL3	XTAL2	XTAL1	XTAL0
TESTCNTRI	25	ISSIN	SMSL	SSTSEL	ISSCOFF	DEMOFF	450LOFF	TESTOU T	TESTIN
TEST	26	- 30	-	-	-	-	DIV2	DIV1	DIV0
TEST MODE1	27	OUT7	OUT6	OUT5	OUT4	OUT3	OUT2	OUT1	OUT0
TEST MODE2	28	-	-	TINMP	TINAC	OUT11	OUT10	OUT9	OUT8

2.2 Control Register Function

Table 2.

Register Name	Function						
A Charge pump high current							
ACF	Adjacent channel filter select						
ACG	Adjacent channel filter gain						
ACM	Threshold for startpoint adjacent channel mute						
ACNTH	Adjacent channel narrow band threshold						
ACWTH	Adjacent channel wide band threshold						
AISS	ISS filter fine adjust						

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Table 2. (continued)

Register Name	Function
AMAGC	AM wide band AGC threshold
AMD	AM prescaler
AMON	AM-FM switch
AMSEEK	Switch time constant for AM seek
AMSS	AM seek stop threshold
AMST	AM stereo select
APPM	Application mode quality detection
В	Charge pump low current
CF	Center frequency IF counter
CLKSEP	Clock separation (only for testing)
CURRH	Set current high charge pump
D	Inlock phase error and delay time for lock detector
DAGC	AM narrow band AGC threshold
DEM	Demodulator offset
DEMOFF	Demodulator clock "OFF" (only for testing)
DNB	Demodulator noise blanking
DIV	Divider ratio for reference frequency (only for feeling)
DTH	Deviation detector threshold for ISS filter CrFr
DWTH	Deviation detector threshold for 193 filter narrow/wide
EW	Frequency error window IF counter
FMAGC	FM AGC threshold
IF1T	FM/AM mixer1 tarts adjust
IF2T	AM mixer2 lank adjust
IFCM	IF counter mode
IFENA	IF courser enable
IFG	, I⊢ î amplifier gain (10.7MHz)
IFS	IF counter sampling time
IRC	IF reference counter
ISSCOFI	ISS filter clock "OFF" (only for testing)
เรกอาบ	ISS filter control
ISilin	Test input for ISS filter
ISSON	ISS filter "ON"
ISS30	ISS filter 30KHz weather band
ISS80	ISS filter narrow/mid switch
LDENA	Lock detector enable
LPF	Loop filter input select
MENA	Softmute enable
MPAC	Adjacent channel control by multipath
MPOFF	Multipath control "OFF"
MPF	Multipath filter frequency
MPG	Multipath filter gain
MPTH	Multipath threshold
OUT	Test output (only for testing)

Table 2. (continued)

Register Name	Function
PC	Counter for PLL (VCO frequency)
RC	Reference counter PLL
RES	Reservation
SL	S meter slider threshold
SLOPE	Softmute slope select
SMCTH	Softmute capacitor threshold for ISS "ON"
SMD	Softmute depth threshold
SMSL	S meter slope
SMTH	Softmute startpoint threshold
SSTSEL	Search stop select for continuous signal
TEST3	Testing PLL/IFC (only for testing)
TESTOUT	Switch FSW output to TEST output (only for testing)
TESTIN	Switch FSU input to TEST input (only for testing)
TDEV	Time constant for deviation detector
TINAC	Test input adjacent channel (only for testing)
TINMP	Test input multipath(only for testing)
TISS	Time constant for ISS filter "ON"/"OFF"
TVR	Tuning voltage for prestage proportional referred to PLL
TVO	Tuning voltage offset for prestage
XTAL	Xtal frequency adjust
450LOFF	450kHz limiter "OFF (only for testing)
Subaddress	inction obsor
Table 3.	

Subaddress

Table 3.

MSB			0		LSB		LSB	Function
T2	T1		.4	A3	A2	A1	A0	Function
		KC	0	0	0	0	0	Charge pump control
			0	0	0	0	1	PLL lock detector
C	O^{-}		-	Ð	-	-	-	-
	1		1	1	1	0	0	Test mode 2
		0						Page mode "OFF"
	10	1	<i>v</i>					Page mode enable

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Note: 1. T1, T2 used for testing, in application mode they have to be "0".

2.3 Data Byte Specification

Addr 0 Charge Pump Control

Table 4.

MSB							LSB	Function
d7	d6	d5	d4	d3	d2	d1	d0	Function
				0	0	0	0	High current = 0mA
				0	0	0	1	High current = 0.5mA
				0	0	1	0	High current = 1mA
				0	0	1	1	High current = 1.5mA
				-	-	-	-	-
				1	1	1	1	High current = 7.5mA
		0	0					Low current = 0µA
		0	1					Low current = 50µA
		1	0					Low current = 100µA
		1	1					Low current = 150µA
	0							Select low current
	1							Select high current
0								Select LPFM
1								Select LP.M.

Addr 1 PLL Lock Detector

Table 5.

1 Addr Table		Lock	Deteo	ctor		10		Select LP.M.	
MSB						77-	LSB	Euro	ction
d7	d6	d5	d4	d3	a.?	d1	d0		Clion
						0	0	Not used, have to be 0, d1 only	/ for testing
			5		0			Select FM mode	
					1	*(וכ	Select AM mode	
		×C	0	0			b	PD phase difference threshold	10ns
	10		0	1					20ns
	07		1	0	J.				30ns
20			1	1					40ns
P	0	0						Not valid	
	0	1						Activation delay	4 x 1/f _{REF}
	1	0							6 x 1/f _{REF}
S	Y	1							8 x 1/f _{REF}
0								Lock detector doesn't control c	harge pump
1								Lock detector controls charge p	oump

Addr 2 PLL Counter 1 (LSB)

Table 6.

MSB							LSB	Function
d7	d6	d5	d4	d3	d2	d1	d0	Function
0	0	0	0	0	0	0	0	LSB = 0
0	0	0	0	0	0	0	1	LSB = 1
0	0	0	0	0	0	1	0	LSB = 2
-	-	-	-	-	-	-	-	-
1	1	1	1	1	1	0	0	LSB = 252
1	1	1	1	1	1	0	1	LSB = 253
1	1	1	1	1	1	1	0	LSB = 254
1	1	1	1	1	1	1	1	LSB = 255

Addr 3 PLL Counter 2 (MSB)

1	1	1	1	1	1	1	1	LSB = 255
Addr Table		Cour	nter 2	(MSB))			produces
MSB							LSB	Function
d7	d6	d5	d4	d3	d2	d1	d0	Gunction
0	0	0	0	0	0	0	0	MSB = 0
0	0	0	0	0	0	0	1	MSB = 250
0	0	0	0	0	0	1	0	MS2 = 512
-	-	-	-	-	-	-	-	-U' x0
1	1	1	1	1	1	0	0	MSB = 64768
1	1	1	1	1	1	0		MSB = 65024
1	1	1	1	1	1	1	0	MSB = 65280
1	1	1	1	1	1	\mathcal{D}_1	1	MSB = 65536

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Note: 1. Swallow mode: $f_{VCO}/f_{S'N} = \frac{1}{S} + MSB + 32$

Addr 4 PLL Reference Counter 1 (LSB)

Table 8.

Table	8.	3,6			(
MSB		5		~			LSB	Function
ďĩ	-26	d5	d4	d3	d2	d1	d0	Function
\mathbf{O}	0	0	0	0	0	0	0	LSB = 0
0	0	0	0	0	0	0	1	LSB = 1
0	0	0	0	0	0	1	0	LSB = 2
	C,	-	-	-	-	-	-	-
	1	1	1	1	1	0	0	LSB = 252
1	1	1	1	1	1	0	1	LSB = 253
1	1	1	1	1	1	1	0	LSB = 254
1	1	1	1	1	1	1	1	LSB = 255

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Addr 5 PLL Reference Counter 2 (MSB)

Table 9.

MSB							LSB	Function
d7	d6	d5	d4	d3	d2	d1	d0	Function
0	0	0	0	0	0	0	0	MSB = 0
0	0	0	0	0	0	0	1	MSB = 256
0	0	0	0	0	0	1	0	MSB = 512
-	-	-	-	-	-	-	-	-
1	1	1	1	1	1	0	0	MSB = 64768
1	1	1	1	1	1	0	1	MSB = 65024
1	1	1	1	1	1	1	0	MSB = 65280
1	1	1	1	1	1	1	1	MSB = 65536

Note: 1. $f_{OSC}/f_{REF} = LSB + MSB + 1$

Addr 6, Addr7, Addr 8 TV1,2 (proportional and offset control referred to tuning voltage V28)

Table 10.

ad	dr8				ade	dr6				Function TV1
d7	d6	d7	d6	d5	d4	d3	d2	d1	d0	Function IV
ad	dr8				ade	dr7				Function TV2
d5	d4	d7	d6	d5	d4	d3	d2	d1	d0	
				0	0	0	0	0	U	TVR = 0
				0	0	0	0	0	1	TVR = -1
				0	0	0	0		0	TVR = -2
-	-	-	-	-	-					05
				0	1	$\mathbf{D}_{\mathbf{i}}$	1	1	1	TVR = -31
				1	U U	0	0	0	1	TVR = +1
				1	0	0	0	1	0	TVR = +2
-	-	-	2-	-	-		Z	-	-	-
			9	1	1	1	1	1	1	TVR = +31
0	Û	5	0		5					TVO = 0
0	50	0	1	20						TVO = -1
0	0	1	0							TVO = -2
-	-	X	0	-	-	-	-	-	-	-
0	1	1	1							TVO = -7
1	0	0	1							TVO = +1
1	0	1	0							TVO = +2
Y	-	-	-	-		-	-	-	-	-
1	1	1	1							TVO = +7

Note: 1. TV1,2 = V28 + V28 * TVR/128 + TVO * 50mV TVR: -31, -30, ... 0 ... 30, 31 TVO: -7, -6, ... 0 ... 6, 7

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Addr 8 IF Counter Control 1

Table 11.

MSB							LSB	Function
d7	d6	d5	d4	d3	d2	d1	d0	Function
					0	0	0	Not valid
					0	0	1	Not valid
					0	1	0	Not valid
					0	1	1	$\Delta f = 6.25 \text{kHz} (FM)1 \text{kHz} (AM UPC)$
					1	0	0	$\Delta f = 12.5 \text{kHz}$ (FM) 2kHz (AM UPC)
					1	0	1	$\Delta f = 25 \text{kHz}$ (FM) 4kHz (AM UPC)
					1	1	0	$\Delta f = 50 \text{kHz}$ (FM) 8kHz (AM UPC)
					1	1	1	Δf = 100kHz (FM)16kHz (AM UPC)
				0				IF counter disable / stand by
				1				IF counter enable
	0 0 0 0 0 1 0 1 1 0 1 1 1 1 1 1 1 1 1 1 1 1 1 1 0 1							ete Product(S)
MSB							LSB	Function

Addr 9 IF Counter Control 2

MSB							LSB	Function
d7	d6	d5	d4	d3	d2	d1	d0	Function
			0	0	0	0	0	fr.ente 10.60625MHz (FM)10.689MHz (AM UPC)
			0	0	0	0	1	f _{Cemer} = 10.61250MHz (FM)10.690MHz (AM UPC)
-	-	-	-	-	-		<u> </u>	-
			0	1	0	<u>, -</u>	00	f _{Center} = 10.66875MHz (FM)10.699MHz (AM UPC)
			0	1	0		1	f _{Center} = 10.67500MHz (FM)10.700MHz (AM UPC)
			0	1	1	0	0	f _{Center} = 10.68125MHz (FM)10.701MHz (AM UPC)
			0		1	0	1	f _{Center} = 10.68750MHz (FM)10.702MHz (AM UPC)
			C	1	1	1	0	f _{Center} = 10.69375MHz (FM)10.703MHz (AM UPC)
		20	Û	1	1	1	1	f _{Center} = 10.70000MHz (FM)10.704MHz (AM UPC)
	10		1	0	0	0	0	f _{Center} = 10.70625MHz (FM)10.705MHz (AM UPC)
	\odot		1	0	0	0	1	f _{Center} = 10.71250MHz (FM)10.706MHz (AM UPC)
0.5	-	-	\mathbf{O}	-	-	-	-	-
			1	1	1	1	1	f _{Center} = 10.80000MHz (FM)10.720MHz (AM UPC)
0	0	0						t _{Sample} = 20.48ms (FM)128ms (AM UPC)
0	0	1						t _{Sample} = 10.24ms (FM)64ms (AM UPC)
0	1	0						t _{Sample} = 5.12ms (FM)32ms (AM UPC)
0	1	1						t _{Sample} = 2.56ms (FM)16ms (AM UPC)
1	0	0						t _{Sample} = 1.28ms (FM)8ms (AM UPC)
1	0	1						t _{Sample} = 640μs (FM)4ms (AM UPC)
1	1	0						t _{Sample} = 320μs (FM)2ms (AM UPC)
1	1	1						t _{Sample} = 160μs (FM)1ms (AM UPC)

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Addr 10 IF Counter Reference (LSB)

Table 13.

MSB							LSB	Function
d7	d6	d5	d4	d3	d2	d1	d0	Function
0	0	0	0	0	0	0	0	LSB = 0
0	0	0	0	0	0	0	1	LSB = 1
0	0	0	0	0	0	1	0	LSB = 2
-	-	-	-	-	-	-	-	-
1	1	1	1	1	1	0	0	LSB = 252
1	1	1	1	1	1	0	1	LSB = 253
1	1	1	1	1	1	1	0	LSB = 254
1	1	1	1	1	1	1	1	LSB = 255

Addr 11 IF Counter Reference (MSB) and IF Counter Mode Select

Table 14	4.
----------	----

MSB			-	-			LSB	Function
d7	d6	d5	d4	d3	d2	d1	d0	
		0	0	0	0	0	0	MSB = 0
		0	0	0	0	0	1	MSB = 25 o
		0	0	0	0	1	0	MGE = 512
		-	-	-	-	-	-	
		1	1	1	1	0	_1	MSB = 15616
		1	1	1	1		29	MSB = 15872
		1	1	1	1	1	1	MSB = 16128
0	0					<u> </u>		Not valid
0	1							IF counter FM mode
1	0						5)	Not valid
1	1							IF counter AM upconversion mode
			SB + MS		20.			

Addr 12 IF 1 and FM AGC

Table 15.

MSB							LSB	Function
d7	d6	d5	d4	d3	d2	d1	d0	Function
						0	0	IF1 gain1 9dB
						0	1	IF1 gain1 11dB
						1	0	IF1 gain1 13dB
						1	1	IF1 gain1 15dB
				0	0			IF1 gain2 9dB
				0	1			IF1 gain2 11dB
				1	0			IF1 gain2 13dB
				1	1			IF1 gain2 15dB
	0	0	0					AGC threshold 80dBµV
	0	0	1					AGC threshold 82dBµV
	0	1	0					AGC threshold 84dBµV
	0	1	1					AGC threshold 86dBµV
	1	0	0					AGC threshold 88dB, V
	1	0	1					AGC threshold 20.15+V
	1	1	0					AGC threshold \$2dBµV
	1	1	1					Keving AGC OFF"
0								has to be "0"
		modu	lator	Fine A	Adjust		3	obsolett
Table								0050

Addr 13 Demodulator Fine Adjust

Table 16.

MSB							LSB	Function
d7	d6	d5	d4	0.7	d2	d1	d0	Function
		0	0	υ	0	0	0	0mV
		0	0	0	0	0	1	+6mV
		0	0	0	0	1	0	+12mV
		0-	-		-	-	-	-
C		0	1	1	1	1	1	+186mV
		1	0	0	0	0	0	0mV
		1	0	0	0	0	1	-6mV
		1	0	0	0	1	0	-12mV
	·O,	-	-	-	-	-	-	-
-10-	2	1	1	1	1	1	1	-186mV
1	1							have to be "1"

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Addr 14 Quality Detection Adjacent Channel

Table 17.

MSB							LSB	Function					
d7	d6	d5	d4	d3	d2	d1	d0	Function					
							1	ISS filter 30KHz "ON"for weather band					
						0		AC highpass frequency 100kHz					
						1		AC bandpass frequency 100kHz					
					0			AC gain 32dB					
					1			AC gain 38dB					
		0	0	0				AC wide band threshold 0.25V					
		0	0	1				AC wide band threshold 0.35V					
		0	1	0				AC wide band threshold 0.45V					
		-	-	-									
		1	1	1				AC wide band threshold 0.95V					
0	0							AC narrow band threshold 0.0%					
0	1							AC narrow band threshold 0. V					
1	0							AC narrow band thre: hv/a 0.2V					
1	1							AC narrow band threshold 0.3V					
	Addr 15 Quality Detection Multipath												
MSB							L3B	Function					

Addr 15 Quality Detection Multipath

MSB							L3B	F irmedian
d7	d6	d5	d4	d3	d2	a 1	du	Function
						b^{-1}	0	Multipath control "ON"
				C			1	Multipath control "OFF"
				\mathbf{O}		0	1	MP bandpass frequency 19KHz
						10	51	MP bandpass frequency 31KHz
		10			0	1	•	MP gain 12dB
	10				1			MP gain 23dB
			0	0	5			MP threshold 0.50V
			0	1				MP threshold 0.75V
JQ.		0	1	0				MP threshold 1.00V
		XC	1	1				MP threshold 1.25V
	0	0						Application mode 1
C	0	1						Application mode 2
0								Multipath eliminates ac
1								Multipath eliminates ac and ac+

Addr 16 Quality Deviation Detection

Table 19.

MSB							LSB	Function
d7	d6	d5	d4	d3	d2	d1	d0	Function
					0	0	0	charge current $34\mu A$, discharge current $6\mu A$
					0	0	1	charge current 32μA, discharge current 8μA
					0	1	0	charge current 30μA, discharge current 10μA
					0	1	1	charge current 28μA, discharge current 12μA
					-	-	-	-
					1	1	1	charge current 20µA, discharge current 20µA
			0	0				DEV threshold for ISS narrow/wide 30kHz
			0	1				DEV threshold for ISS narrow/wide 45kHz
			1	0				DEV threshold for ISS narrow/wide 60kHz
			1	1				DEV threshold for ISS narrow/wide 75I Hz
	0	0						DEV threshold for ISS filter "OFF' retic 1
	0	1						DEV threshold for ISS filter "CrF" ratio 1.3
	1	0						DEV threshold for ISS filter "OFF" ratio 1.4
	1	1						DEV threshold for ISS niter "OFF" ratio 1.5
0								has to be 0

Addr 17 Quality ISS Filter

Table 20.

0								has to be 0
Addr Table		iality I	ISS Fil	lter				ObstetePi
MSB							J.SB	c0
d7	d6	d5	d4	d3	d2	- -1	d0	Function
							0	ISS filter control "ON"
				0	D -		1	ISS filter control "OFF"
						0		Switch ISS filter "OFF"
						1		Switch ISS filter "ON"
		KO			0	0-	[Switch ISS filter 120kHz
		\overline{D}			1			Switch ISS filter 80kHz
C		0	0	0				discharge current1 μ A, charge current mid 74 μ A narrow124 μ
0-		0	0	1				discharge current3µA, charge current mid 72µA narrow122µ
		0	1	0				discharge current5µA, charge current mid 70µA narrow120µ
		0	1	1				discharge current7µA, charge current mid 68µA narrow118µ
C	О,	-	-	-				-
5		1	1	1				discharge current15µA,charge current mid 60µAnarrow110µ
0	0							ISS filter fine adjust -20kHz
0	1							ISS filter fine adjust -10kHz
1	0							ISS filter fine adjust 0kHz
1	1							ISS filter fine adjust +10kHz

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Addr 18 AM Control1

Table 21.

MSB							LSB	Function
d7	d6	d5	d4	d3	d2	d1	d0	
							0	Normal AGC time constant
							1	Short time constant for AM seek stop
						0		Multipath information available FM
						1		AM stereo output available
				0	0			Prescaler ratio 10
				0	1			Prescaler ratio 8
				1	0			Prescaler ratio 6
				1	1			Prescaler ratio 4
0	0	0	0					Narrow band AGC threshold 74.4dBµV
0	0	0	1					Narrow band AGC threshold 78.8dBµV
0	0	1	0					Narrow band AGC threshold 80.0dLu
0	0	1	1					Narrow band AGC threshold SJ.7dBµV
0	1	0	0					Narrow band AGC three hold 53.2dBµV
0	1	0	1					Narrow band ACC #resnold 77.1dBµV
0	1	1	0					Narrow band AGC threshold 78.5dBµV
0	1	1	1					Narro , pand AGC threshold 79.4dBµV
1	0	0	0					Narro v band AGC threshold 42.7dBμV
1	0	0	1					Narrow band AGC threshold 65.8dBµV
1	0	1	0			10		Narrow band AGC threshold 77.6dBµV
1	0	1	1					Narrow band AGC threshold 78.5dBµV
1	1	0	0			7	Ì	Narrow band AGC threshold 32.6dBµV
1	1	0	1		50-			Narrow band AGC threshold 55.0dBµV
1	1	1	0	$[\mathbf{O}]$				Narrow band AGC threshold 73.3dBµV
1	1	1	1			X	51	Narrow band AGC threshold 77.6dBµV

1.	мсв		0					LSB	Function
Γ	d7	d6	d5	d4	d3	d2	d1	d0	Function
Γ			0		0	0	0	0	AGC Threshold 82.0dBµV 80.1dBµV
Γ	0	5			0	0	0	1	AGC Threshold 85.4dBµV 83.4dBµV
K	0				0	0	1	0	AGC Threshold 87.5dBµV 85.5dBµV
					0	0	1	1	AGC Threshold 89.2dBµV 87.2dBµV
					0	1	0	0	AGC Threshold 90.6dBµV 88.6dBµV
					0	1	0	1	AGC Threshold 91.9dBµV 89.8dBµV
					0	1	1	0	AGC Threshold 92.9dBµV 90.8dBµV
					0	1	1	1	AGC Threshold 93.8dBµV 91.8dBµV
					1	0	0	0	AGC Threshold 94.6dBµV 92.6dBµV
ſ					1	0	0	1	AGC Threshold 95.4dBµV 93.3dBµV
l					1	0	1	0	AGC Threshold 96.1dBµV 94.0dBµV

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Table 22. (continued)

MSB							LSB	Function
d7	d6	d5	d4	d3	d2	d1	d0	Function
				1	0	1	1	AGC Threshold 96.7dBµV 94.6dBµV
				1	1	0	0	AGC Threshold 97.3dBµV 95.2dBµV
				1	1	0	1	AGC Threshold 97.8dBµV 95.7dBµV
				1	1	1	0	AGC Threshold 98.4dBµV 96.3dBµV
				1	1	1	1	AGC Threshold 98.8dBµV 96.8dBµV
0	0	0	0					Seek stop threshold 20.3dBµV
0	0	0	1					Seek stop threshold 20.8dBµV
0	0	1	0					Seek stop threshold 21.3dBµV
0	0	1	1					Seek stop threshold 22.0dBµV
0	1	0	0					Seek stop threshold 22.8dBµV
0	1	0	1					Seek stop threshold 23.7dBµV
0	1	1	0					Seek stop threshold 24.7dBµV
0	1	1	1					Seek stop threshold 25.9dBµV
1	0	0	0					Seek stop threshold 27.0aBµV
1	0	0	1					Seek stop thres or 28.3dBµV
1	0	1	0					Seek stop thiss old 30.4dBµV
1	0	1	1					Seek Stop threshold 32.5dBµV
1	1	0	0					Seek stop threshold 35.2dBµV
1	1	0	1					Seek stop threshold 37.6dBµV
1	1	1	0			10		Seek stop threshold 40.2dBµV
1	1	1	1					Seek stop threshold 43.0dBµV
Addr : Fable		ftmute	e Con	trol		ر	5)	, OV
MSB		.0			. (17-	LSB	Function

Ī	MSB					. (ŤŢ.	LSB	F
Î	d7	d6	35	d4	d3	d2	d1	d0	Function
Ĩ		67			0	0	0	0	Mute depth 0 in application 18dB
	3			0	0	0	0	1	Mute depth 1 in application 20dB
			. 0		0	0	1	0	Mute depth 2 in application 22dB
			XC		0	0	1	1	Mute depth 3 in application 24dB
Ī			0		-	-	-	-	- (logarithmically behaviour)
	C	Ď.			1	1	1	1	Mute depth 15 in application 36dB
	0-			0					Mute disable
				1					Mute enable
			0						Sharp slope
Ī			1						Smooth slope
Î	0	0							Function "OFF"
Î	0	1							Mute depth threshold for ISS filter "ON" 2dB
Î	1	0							Mute depth threshold for ISS filter "ON" 1dB
Î	1	1							Mute depth threshold for ISS filter "ON" 0.2dB



Addr 21 Softmute Control 2

Table 24.

MSB							LSB	Function
d7	d6	d5	d4	d3	d2	d1	d0	Function
				0	0	0	0	Startpoint mute 0 in application 3dBµV
				0	0	0	1	Startpoint mute 1 in application 4dBµV
				0	0	1	0	Startpoint mute 2 in application 5dBµV
				0	1	0	0	Startpoint mute 3 in application 6dBµV
				-	-	-	-	-
				1	1	1	1	Startpoint mute 15 in application 18dBµV
0	0	0	0					AC mute threshold 60mV
0	0	0	1					AC mute threshold 80mV
0	0	1	1					AC mute threshold 100mV
-	-	-	0					·
1	0	0	0					AC mute threshold 220mV
1	1	1	1					AC mute "OFF"
Addr∶ Table		Neter	Slider					solete product
MSB							LSB	Function
d7	d6	d5	d4	d3	d2	d1	0b	

Addr 22 S-Meter Slider

MSB							LSB	Function
d7	d6	d5	d4	d3	d2	d1	d0	Function
		0	0	0	0	0	0	S meter offset 0mV
		0	0	0	0	7	51	S meter offset -58mV
		0	0	0	0	1	0	S meter offset -116mV
		-	-	-		-	-	-0+
		0	1		1	1	1	S meter offset -1798mV
		1	0	Ŭ	0	0	0	S meter offset 0mV
		1	3	0	0	0	1	S meter offset +58mV
			0	0	0	J1	0	S meter offset +116mV
		7 -	-	-)-	-	-	-
G		1	1	1	1	1	1	S meter offset +1798mV
	х		X					Not used
005 005		ste						

Addr 23 Tank Adjust

Table 26.

d7	d6						LSB	Eurotion
		d5	d4	d3	d2	d1	d0	Function
				0	0	0	0	450kHz 0pF
				0	0	0	1	450kHz 1.6pF
				0	0	1	0	450kHz 3.2pF
				0	0	1	1	450kHz 4.8pF
				-	-	-	-	-
				1	1	1	1	450kHz 24pF
0	0	0	0					10.7MHz 0pF
0	0	0	1					10.7MHz 0.55pF
0	0	1	0					10.7MHz 1.1pF
0	0	1	1					10.7MHz 1.65pF
-	-	-	-					-
1	1	1	1					10.7MHz 8.25pF
Addr 2 Table		AL Ac	ljust					solete product
MSB							LSB	Function

Addr 24 XTAL Adjust

Table 27.

MSB							LSB	Function
d7	d6	d5	d4	d3	d2	d1	d0	Function
			0	0	0	0	0	CLoad 0pF
			0	0	0) (51	C _{Load} 0.625pF
			0	0	0	77	0	C _{Load} 1.25pF
			0	0	0	1	1	C _{Load} 1.875pF
			0	U	1	0	0	C _{Load} 2.5pF
			\mathbf{O}	- -	-	. + C	5	-
				1	1	1	1	C _{Load} 19.4pF
		0	1					Only for testing has to be "0"
х	X)			Not used

ddr 25 Test Control،

Table 28.

Ī	MSB	01						LSB	Function	
	d7	d6	d5	d4	d3	d2	d1	d0	Function	
	Y			0	0	0	0	0	Only for testing (bits have to be "0")	
			0						AMIF select	
Ī			1						Permanent search stop select for AM	
Î		0							S meter slope 1V/decade	
Î		1							S meter slope 1.5V/decade	
	0								Only for testing (has to be "0")	

Addr 26 Test

Table 29.

MSB							LSB	Function			
d7	d6	d5	d4	d3	d2	d1	d0	Function			
х	х	х	х	х	1	1	0	Only for testing (bits have to be set)			

Addr 27 Test Mode1

d7 0 Addr 2 Table MSB d7
Addr 2 Fable MSB
Table MSB
d7
х
05

MSB							LSB	Function
d7	d6	d5	d4	d3	d2	d1	d0	Gunction
х	х	0	0	0	0	0	0	Only for testing (bi s rave to be set)

APPENDIX

Figure 2.



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Block Diagram Quality Detection Principle

Table 32.

Signal	LOW	HIGH
ac	No adjacent channel	Adjacent channel present
ac+	No strong adjacent channel	Adjacent channel higher as ac
sm	Fieldstrength higher as softmute threshold	Fieldstrength lower as softmute threshold
dev	Deviation lower as threshold DWTH	Deviation higher as threshold DWTH
dev+	Deviation lower as threshold DTH*DWTH	Deviation higher as threshold DTH*DWTH
inton	ISS filter off by logic (wide)	ISS filter on by logic
int80	ISS filter 120kHz (mid)	ISS filter 80kHz (narrow)

Table 33.

	Ir	nput Signa	ls			Mode1			Mode?	
ac	ac+	sm	dev	dev+	inton	int80	Function	inton	in 181)	Function
0	0	0	0	0	0	0	wide	0	0	wide
0	0	0	1	0	0	0	wide	0	0	wide
0	0	0	1	1	0	0	wide		0	wide
0	0	1	0	0	1	1	narrow	1	1	narrow
0	0	1	1	0	0	0	Vide	1	0	mid
0	0	1	1	1	0	0	wide	0	0	wide
1	0	0	0	0	1	1	narrow	1	0	mid
1	1	0	0	0	1	51	narrow	1	1	narrow
1	0	0	1	0	1	0	mid	1	0	mid
1	1	0	1	1	1	0	mid	1	1	narrow
1	0	1	0	C	1	1	narrow	1	1	narrow
1	1	1	0		1	1	narrow	1	1	narrow
1	0	1	1	5	1	0	mid	1	0	mid
1	1	1		0	1	0	mid	1	1	narrow
1	0	1	1	1		0	mid	1	0	mid
1	1	1	1	1	1	0	mid	1	1	narrow

Part List (Application- and Measurment Circuit)

0

Table 34.

item	Description
	TOKO 7KL 600ENS-9132NK
152	TOKO 7KL 600ENS-9390AQ
F3	TOKO 7KL V600ENS-9469BS
F4	TOKO 7PSG 628AC-5022N
F5	TOKO PGL 5PGLC-5103N
L1,L5	SIEMENS SIMID03 B82432 1mH
L2,L4	TOKO FSLU 2520-680 68µH
L3	TOKO FSLU 2520-150 15µH
CF1,CF2	MURATA SFE10.7MS3A10-A 180KHz or (TOKO CFSK107M3-AE-20X)
CF3	MURATA SFE10.7MJA10-A 150KHz or (TOKO CFSK107M4-AE-20X)
CF4	MURATA SFPS 450H 6KHz or (TOKO ARLFC450T)
T1,T2	COMPONEX B4F 617PT-1026
D1,D3,D5	TOKO KV1410
D2,D4	TOSHIBA 1SV172
Q1	TOSHIBA 3SK126
Q2	TOSHIBA HN3G01J

3 APPLICATION CIRCUI

Figure 4.



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MEASUREMENT CIRCUIT 4

Figure 5.



DIM.		mm		inch			
0	MIN.	TYP.	MAX.	MIN.	TYP.	MAX.	
А			1.60			0.063	
A1	0.05		0.15	0.002		0.006	
A2	1.35	1.40	1.45	0.053	0.055	0.057	
В	0.18	0.23	0.28	0.007	0.009	0.011	
С	0.12	0.16	0.20	0.0047	0.0063	0.0079	
D		12.00			0.472		
D1		10.00			0.394		
D3		7.50			0.295		
е		0.50			0.0197		
E		12.00			0.472		
E1		10.00			0.394		
E3		7.50			0.295		
L	0.40	0.60	0.75	0.0157	0.0236	0.0295	
L1		1.00			0.0393		
К		()°(min.),	7°(max.)		

OUTLINE AND MECHANICAL DATA TQFP64



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